## **EE 438**

Exam No. 3

Spring 2002

- You have 50 minutes to work the following four problems.
- Be sure to show all your work to obtain full credit.
- The exam is closed book and closed notes.
- Calculators are permitted.
- 1. (30 pts.)
  - a. A wide-sense stationary process X[n] with mean 1 and autocorrelation function

$$r_{XX}[n] = \begin{cases} 1, & n = 0 \\ 0.5, & n = \pm 1 \\ 0, & \text{else} \end{cases}$$

is input to the system described by y[n] = x[n] - x[n-1].

Find the mean  $\mu_{\gamma}[n]$  and autocorrelation  $r_{\gamma\gamma}[n]$  of the output process Y[n].

b. The same wide-sense stationary process X[n] as in part a. above is input to the system described by y[n] = x[2n].

Find the mean  $\mu_{\gamma}[n]$  and autocorrelation  $r_{\gamma\gamma}[n]$  of the output process Y[n].

c. The same wide-sense stationary process X[n] as in parts a. and b. above is input to the system described by

$$y[n] = \begin{cases} x[n/2], & n/2 \text{ is an integer} \\ 0, & \text{else} \end{cases}$$

Find the mean  $\mu_{\gamma}[n]$  and autocorrelation  $r_{\gamma\gamma}[n]$  of the output process Y[n].

$$G. M_{\gamma}[n] = E[\gamma[n]]$$

$$= E[\chi[n] - \chi[n-1]]$$

$$= [-1]$$

$$= 0$$

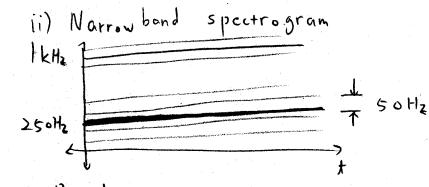
$$\Gamma_{\gamma\gamma}[n] = E[\gamma[m] \gamma[m+n]]$$

$$= E[(\chi[m] - \chi[m-1])(\chi[m+n] - \chi[m+n-1])$$

$$= \Gamma_{\chi\chi}[n] - \Gamma_{\chi\chi}[n-1] - \Gamma_{\chi\chi}[n+1] + \Gamma_{\chi\chi}[n]$$

$$= 2\Gamma_{\chi\chi}[n] - \Gamma_{\chi\chi}[n-1] - \Gamma_{\chi\chi}[n+1] + \Gamma_{\chi\chi}[n]$$

- 2. (20 pts) An individual with pitch frequency 50 Hz utters a voiced phoneme with a strong first formant at 250 Hz and a weaker second formant at 1 kHz.
  - a. Sketch the wideband and narrowband spectrograms corresponding to this speech waveform. Be sure to label all important quantities.
  - b. Design a digital system consisting of a pulse generator driving a linear filter to synthesize this waveform. The system operates at an 8 kHz sampling rate. For this system, specify the pulse interval in samples and the approximate location in the Z plane of the poles for the filter.



b. Pitch period 
$$P = \frac{1}{50} \times 8000 = 160$$
 samples
$$e[n] \rightarrow |V(z)| \frac{S[n]}{\sqrt{A}} \frac{S(t)}{\sqrt{A}}$$

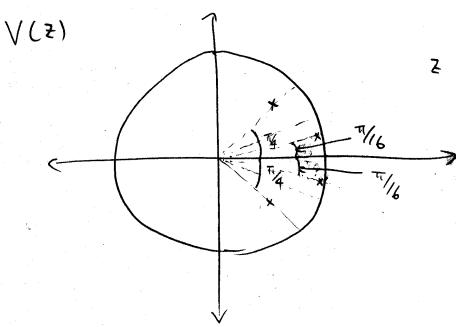
$$f_{s=8} |_{kH_{z}}$$

First formant frequency
$$W_1 = 271 \cdot \frac{250}{8000} = \frac{71}{16}$$

## 2. (continued)

Se cond formant frequency:  

$$\omega_2 = 2\pi \cdot \frac{1000}{8000} = \frac{\pi}{4}$$



3 (25 pts.) In class, we defined the STDTFT of the signal x[n] as

$$X(\omega,n) = \sum_{k} x[k]w[n-k]e^{-j\omega k}.$$

Suppose that we evaluate the STDTFT at L points  $\omega_l = 2\pi l / L, l = 0,..., L-1$  along the frequency axis; so

 $X[l,n] = \sum_{k} x[k]w[n-k]e^{-j2\pi ik/L}$ 

a. Show that for each fixed value of l, X[l,n] can be viewed as the output of a narrowband filter. Find an expression for the frequency response of this filter, and sketch what it would typically look like.

In class, we also showed that x[n] could be reconstructed by summing the outputs of these L filters, provided the impulse responses  $h_l[n]$  of the filters all had value 1/L at n=0, and value 0 at  $n=\pm L, \pm 2L, \pm 3L,...$ 

b. Show that a filter with frequency response

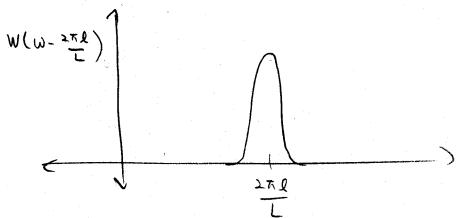
$$H(\omega) = \begin{cases} \frac{1}{L} (1 - |\omega| / (2\pi/L)|), & |\omega| < 2\pi/L \\ 0, & \text{else} \end{cases}$$

satisfies this condition.

$$Q. \times [l, n] = \sum_{k} \sum_{j=1}^{n} \sum_{k} |w[n-k]| e^{-j \frac{2\pi k}{L}}$$

$$= \sum_{k} \sum_{j=1}^{n} \sum_{k} |w[k]| e^{-j \frac{2\pi k}{L}}$$

$$= e^{-j \frac{2\pi k}{L}} \left( \sum_{k} \chi[n-k] |w[k]| e^{-j \frac{2\pi k}{L}} \right)$$



## 3. (continued)

b. 
$$H(\omega) = \left\{ 1 - \left| \frac{\omega}{(2\pi/L)} \right|, |\omega| < 2\pi/L \right\}$$
else

Solution I.

The condition for he [n] for perfect reconstruction is equivalent to

1=0 H(w- 2xl)=1

By the following graph we can see that the condition is satisfied  $H_{L-2}(\omega)$   $H_{L-2}(\omega)$   $H_{L-2}(\omega)$   $H_{L-2}(\omega) = H(\omega - \frac{2\pi l}{L})$ 

Solution II.  $H(\omega) = \int_{\mathbb{R}} \operatorname{rect}\left(\frac{\omega}{2\pi/L}\right) + \int_{\mathbb{R}} \operatorname{rect}\left(\frac{\omega}{2\pi/L}\right).$   $\int_{\mathbb{R}} \operatorname{sinc}\left(\frac{\pi}{L}n\right) \in \frac{\operatorname{DTFT}}{\operatorname{JL}} \int_{\mathbb{R}} \operatorname{rect}\left(\frac{\omega}{2\pi/L}\right).$   $h_{\mathbb{R}}[n] = \int_{\mathbb{R}} \operatorname{sinc}\left(\frac{\pi}{L}n\right) \times \int_{\mathbb{R}} \operatorname{sinc}\left(\frac{\pi}{L}n\right)$   $= \int_{\mathbb{R}} \operatorname{sinc}^{2}\left(\frac{\pi}{L}n\right)$   $h_{\mathbb{R}}[v] = V$   $h_{\mathbb{R}}[k] = 0 \quad \text{where } k \text{ is in integer and } k \neq 0.$ 

4. (25 pts.) Your grades on quizzes 1, 2, and 3 are 3, 4, and 7, respectively. Based on this data set, find the least squares prediction  $\hat{g}[4]$  for your grade on quiz 4, where the predictor has the form  $\hat{g}[n] = a_0 + a_1 n$ . Here n denotes the index of the quiz; and the coefficients  $a_0$  and  $a_1$  are chosen to minimize

$$E = \sum_{n=1}^{3} |\hat{g}[n] - g[n]^{2},$$

where g[n] is the grade for the n-th quiz.

$$E = \sum_{n=1}^{3} \left[ \hat{g}[n] - Q_{0} - Q_{1}n \right]^{2}$$

$$\frac{\partial E}{\partial Q_{0}} = 2 \sum_{n=1}^{3} \left( \hat{g}[n] - Q_{0} - Q_{1}n \right) = 0$$

$$3a_{0} + a_{1}(1+2+3) = 3+4+7$$
  
 $3a_{0} + 6a_{1} = 14$  — 0

$$\frac{\partial E}{\partial a_i} = 2 \sum_{n=1}^{3} (\hat{\beta}[n] - \alpha_0 - \alpha_i n) n = 0$$

$$Q_0(1+2+3) + Q_1(1^2+2^2+3^2) = 3 \cdot 1 + 4 \cdot 2 + 7 \cdot 3$$
  
 $6Q_0 + 14Q_1 = 32$   $-Q_1$ 

From 
$$0 & 0 & 0$$
  
 $0 = \frac{2}{3}$ ,  $0 = 2$ 

$$9[4] = 0.0 + 40.0$$
  
=  $8 + \frac{2}{3}$   
 $\sim 8.67$