

Name: Solution

**General Instructions:**

- You have 60 minutes to complete the exam.
- Write your name on every page of the exam.
- The exam is closed book and closed notes. Calculators are not allowed.
- Your work must be explained to receive full credit. All plots must be carefully drawn with axes labeled.
- Point values for each problem are as indicated. The exam totals 100 points.
- If you finish the exam during the first 50 minutes, you may turn it in and leave. During the last 10 minutes you must remain seated until we pick up exams from everyone.
- Problems labeled with **LO** indicate that the problem is used to determine student satisfaction of course learning objectives.

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This exam is for Krogmeier's section of 301.

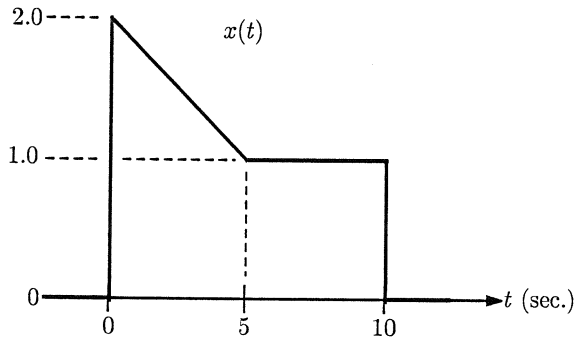
Do not open the exam until you are told to begin.

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Problem 1. *Signal Properties*. [30 pts. total, LO-i]

This problem has three unrelated parts.

(a) [5 pts.] If  $x(t)$  is the signal shown, plot  $y(t) = x(1 - 2t)$ .

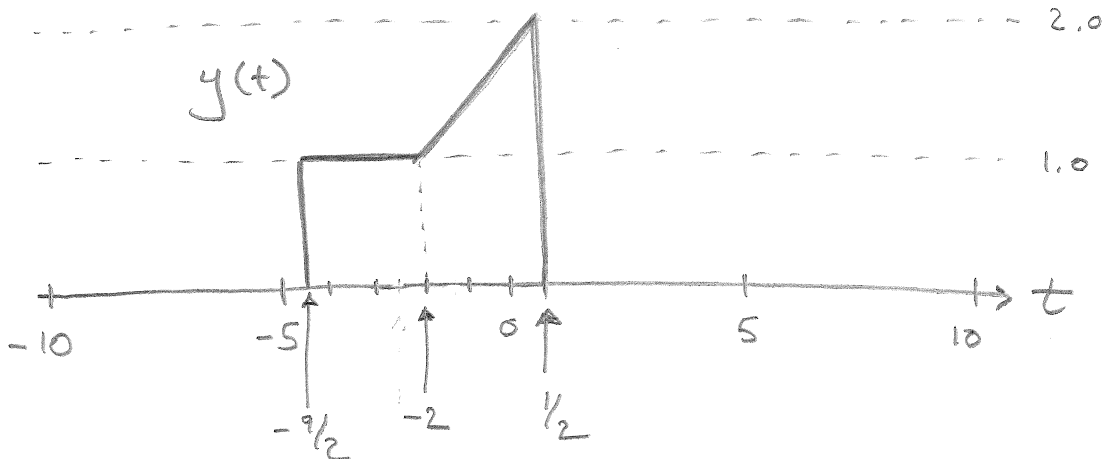


$y(t)$  will be some sort of stretched/compressed and time reversed version of  $x(t)$ . Solve for values of  $t$  st

$$1 - 2t = 0 \longrightarrow t = \frac{1}{2}$$

$$= 5 \longrightarrow -2t = 4 \longrightarrow t = -2$$

$$= 10 \longrightarrow -2t = 9 \longrightarrow t = -\frac{9}{2}$$



(b) [10 pts.] For each of the signals below simplify to the maximum possible extent and find the average power in each.

(b-1)  $x_1(t) = 2 \cos(20\pi t) - 2 \sin(20\pi t)$ . Try to write in form  $A \cos(20\pi t + \theta)$

$$= 2 \operatorname{Re} \left\{ e^{j20\pi t} + j e^{j20\pi t} \right\}$$

$$= 2 \operatorname{Re} \left\{ (1+j) e^{j20\pi t} \right\} = 2 \operatorname{Re} \left\{ \sqrt{2} e^{j\pi/4} e^{j20\pi t} \right\}$$

$$= 2\sqrt{2} \cos(20\pi t + \pi/4)$$

$\therefore$  From 201 average power formula for a sinusoid

$$P = \frac{1}{2} (2\sqrt{2})^2 = 4$$

(b-2)  $x_2(t) = 2 \cos(20\pi t + \pi/4) - 2 \sin(20\pi t + 3\pi/4)$ .

$$= 2 \operatorname{Re} \left\{ e^{j(20\pi t + \pi/4)} + j e^{j(20\pi t + 3\pi/4)} \right\}$$

$$= 2 \operatorname{Re} \left\{ \left( e^{j\pi/4} + j e^{j3\pi/4} \right) e^{j20\pi t} \right\}$$

$$= 2 \operatorname{Re} \left\{ \underbrace{\left( 1 + e^{j\pi/2} e^{j\pi/2} \right)}_{1 + (j)^2 = 1 - 1 = 0} e^{j\pi/4} e^{j20\pi t} \right\}$$

$\therefore x_2(t) = 0 \implies P = 0$

Problem 1. (cont'd.)

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(c) [15 pts.] Let  $w(t) = e^{-3t}u(t)$  and define

$$z(t) = \sum_{k=-\infty}^{\infty} w(t - 10k).$$

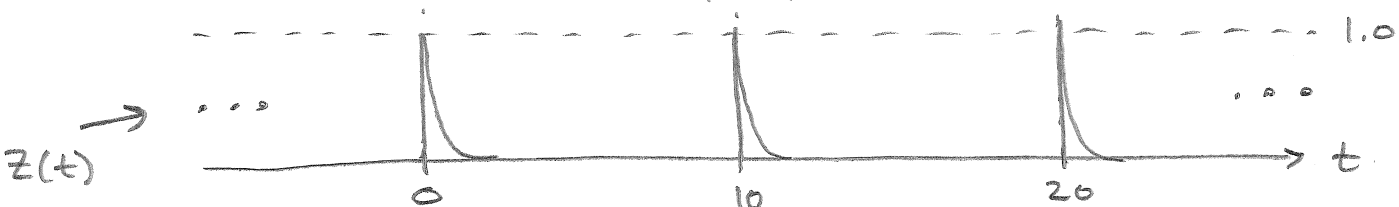
(To simplify calculations be sure to make reasonable engineering approximations. Explain.)

(c-1) Find the energy of  $w(t)$ .

$$E = \int_0^{\infty} w^2(t) dt = \int_0^{\infty} e^{-6t} dt = -\frac{1}{6} e^{-6t} \Big|_{t=0}^{\infty} = \frac{1}{6}$$

(c-2) Is  $z(t)$  periodic or non-periodic? If periodic, find its fundamental period. Make a rough plot using reasonable assumptions. Explain.

$z(t)$  is clearly periodic of period  $T=10$  because of the repeating shifts in the sum. Note that  $w(10) = e^{-30} \approx 0$  so for all intents and purposes there is no overlap:



(c-3) Find the average power of  $z(t)$ .

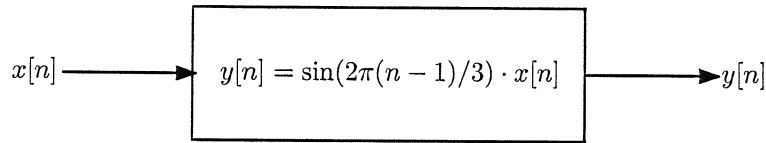
$$P = \frac{\text{energy of } w}{\text{period}} = \frac{1/6}{10} = \frac{1}{60}$$

# Solution

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## Problem 2. System Properties. [40 pts. total, LO-i]

This problem has four parts. For each system shown, choose the correct statements. You must give some explanation of your choices in order to receive full credit.



(a) [10 pts.] The system above is (circle all correct choices, and explain):

(a-1) Memoryless or with memory. Is memoryless since output at a specific time  $n$  only depends on input at the same time  $n$ .  $\sin(2\pi(n-1)/3)$  is just a time-varying gain.

(a-2) Causal or non-causal. Memoryless is always causal.

(a-3) BIBO stable or unstable. Suppose there exists a  $B < \infty$  st.  $|x[n]| \leq B$  for all  $n \in \mathbb{Z}$ . Then

$$|y[n]| = |\sin(2\pi(n-1)/3)| \cdot |x[n]| \leq |x[n]| \leq B \quad \forall n.$$

↑  
because |sinewave| ≤ 1

$$\Rightarrow |y[n]| \leq B \quad \forall n.$$

(a-4) Time invariant or time varying.

Pretty obvious since gain is time varying. Let  $G_n \triangleq \sin(2\pi(n-1)/3)$   
 $\Rightarrow G_1 = 0, G_2 = \sin 2\pi/3, G_3 = \sin 4\pi/3, G_4 = 0 \dots$  is periodic  
 $= \sqrt{3}/2, = -\sqrt{3}/2$

(a-5) Linear or non-linear.

Say  $y_1[n] = G_n x_1[n]$   
 $y_2[n] = G_n x_2[n]$

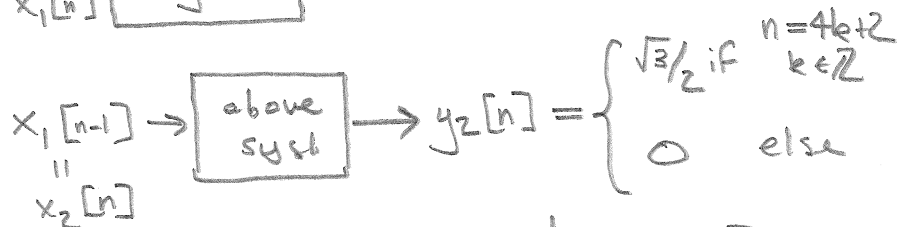
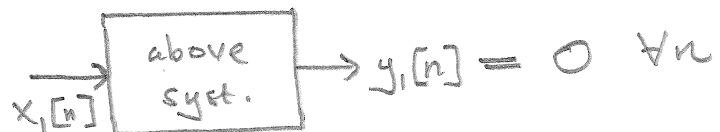
Then

$$G_n (\alpha x_1[n] + x_2[n])$$

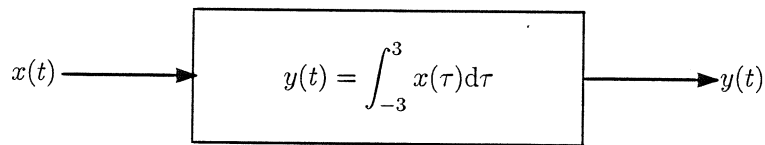
$$= \alpha G_n x_1[n] + G_n x_2[n]$$

$$= \alpha y_1[n] + y_2[n]$$

Let  $x_1[n] = \begin{cases} 1 & \text{if } n=4k+1 \text{ (a mult of 4 plus 1)} \\ 0 & \text{else} \end{cases}$



But  $y_2[n] \neq y_1[n-1]$



(b) [10 pts.] The system above is (circle all correct choices, and explain):

(b-1) Memoryless or **with memory.** Kind of obvious that there is memory since output depends on a range of values  $x(t) : -3 \leq t \leq 3$  of input.

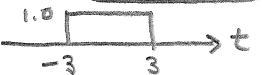
(b-2) Invertible or **non-invertible.** All I have to do is demonstrate a nonzero input signal which results in the identically zero output signal

Consider   $\implies y_1(t) = 0 \forall t \therefore$  system cannot be invertible.

(b-3) **BIBO stable** or unstable. For any particular input the output is a constant (dc) signal. Say  $|x(t)| \leq B \forall t$ . Then

$$|y(t)| = \left| \int_{-3}^3 x(\tau) d\tau \right| \leq \int_{-3}^3 |x(\tau)| d\tau \leq 6B$$

(b-4) Time invariant or **time varying.**

Let   $\implies y_1(t) = 6 \forall t$ . Define  $x_2(t) = x_1(t-6)$

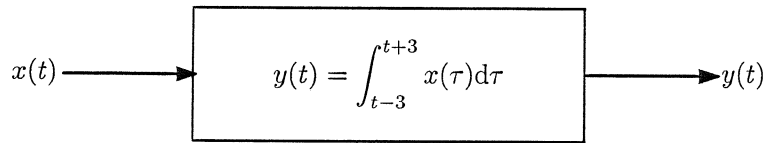
$\hookrightarrow x_2(t) = 0$  for  $-3 < t < 3 \implies y_2(t) = 0 \forall t$ .

Note:

$$y_2(t) \neq y_1(t-6)$$

(b-5) **Linear** or non-linear.

Sums or integrals of signals are always linear.



(c) [10 pts.] The system above is (circle all correct choices, and explain):

(c-1) Memoryless or with memory. output at time  $t$  depends on a range  $x(s) : t-3 < s < t+3$ .

(c-2) Invertible or non-invertible. → Any input which is periodic of period 6 with zero mean will produce the "all zero" output.

(c-3) BIBO stable or unstable. IF carefully check the proof given in b-3 you will see that it goes through.

Suppose  $y_1(t) = \int_{t-3}^{t+3} x_1(\tau) d\tau$ . Then define

(c-4) Time invariant or time varying.  $x_2(t) = x_1(t-T)$  and compute

$$y_2(t) = \int_{t-3}^{t+3} x_2(\tau) d\tau = \int_{t-3}^{t+3} x_1(\tau-T) d\tau = \int_{t-3-T}^{t+3-T} x_1(s) ds = \int_{(t-T)-3}^{(t-T)+3} x_1(s) ds = y_1(t-T)$$

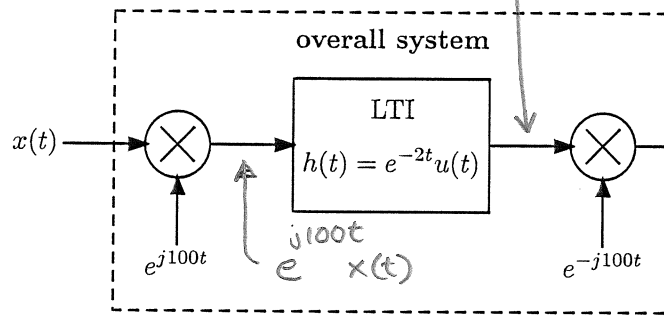
$\uparrow$   $s = \tau - T$

(c-5) Linear or non-linear.

Sums or integrals of signals are always linear.

# Solution

Problem 2. (cont'd.)



$$= \int_0^{\infty} e^{-2\tau} x(t-\tau) e^{j100(t-\tau)} d\tau$$

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$$= e^{j100t} \int_0^{\infty} e^{-(2+j100)\tau} x(t-\tau) d\tau$$

$$= \int_0^{\infty} e^{-(2+j100)\tau} x(t-\tau) d\tau$$

(d) [10 pts.] The system above is (circle all correct choices, and explain):

(d-1) Invertible or non-invertible. To show I need to demonstrate an inverse syst. i.e. find  $h_{inv}(t)$  st.  $h_{overall} * h_{inv}(t) = \delta(t)$

∴ From  $x \mapsto y$  it is LTI with impulse response  $h_{overall}(t) = e^{-(2+j100)t} u(t)$

(d-2) Causal or non-causal. Impulse resp. is causal.

(d-3) BIBO stable or unstable. Easy to show;  $\int_0^{\infty} e^{-(2+j100)t} dt = \int_0^{\infty} e^{-2t} dt < \infty$

(d-4) Time invariant or time varying. That's the "TI" in LTI.

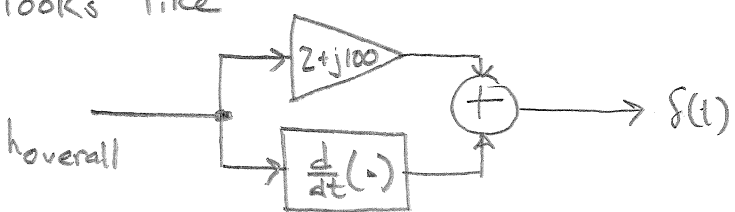
(d-5) Linear or non-linear. The "L" in LTI.

A strategy: Try to find some linear combination of derivatives, etc. which takes  $h_{overall}$  to  $\delta$ .

$$\frac{d}{dt} h_{overall}(t) = -(2+j100) e^{-(2+j100)t} u(t) + e^{-(2+j100)t} \delta(t)$$

$$= -(2+j100) e^{-(2+j100)t} u(t) + \delta(t)$$

⇒  $(2+j100) h_{overall}(t) + \frac{d}{dt} h_{overall}(t) = \delta(t)$  ∴ inverse system looks like



∴  $h_{inv}(t) = (2+j100)\delta(t) + \delta'(t)$



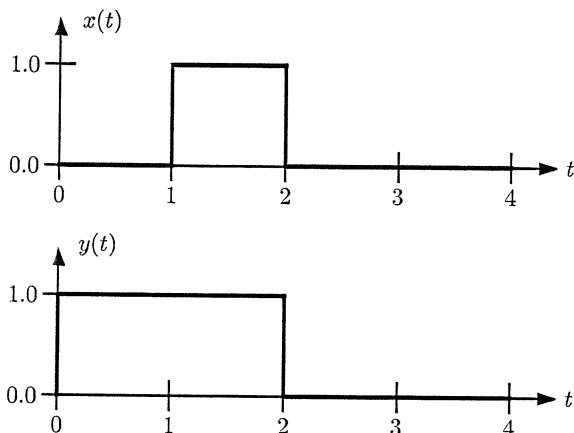
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**Problem 3. Linear and Time-Invariant Systems.** [30 pts. total, LO-ii, LO-iii]

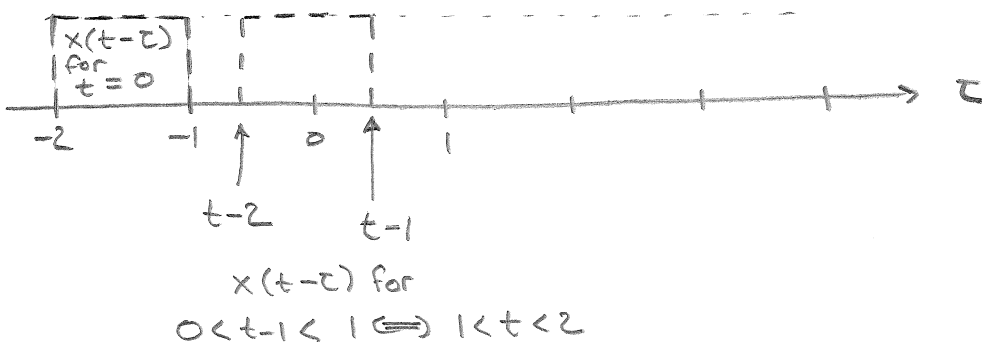
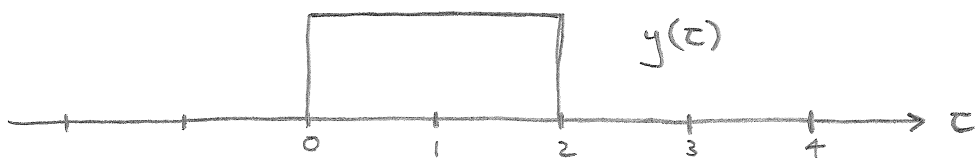
The problem has two parts.

(a) [15 pts.] Compute and plot the convolution of the two rectangular pulses shown below.

Its very helpful to know that conv. rectangular signals yields outputs with ramps in them



$$z(t) = x * y(t) = \int y(\tau) x(t-\tau) d\tau$$



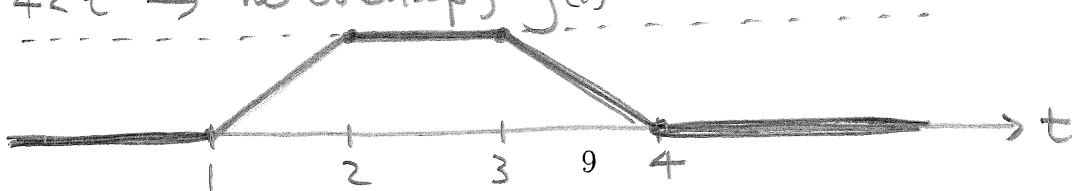
Cases:  $t-1 < 0 \Leftrightarrow t < 1 \Rightarrow$  no overlap;  $y(t) = 0$

$1 < t < 2 \Rightarrow$  partial overlap;  $y(t)$  is linear with pos. slope

$2 < t < 3 \Rightarrow$  full overlap;  $y(t) = \text{const.} = \text{area under } x(t) = 1$

$3 < t < 4 \Rightarrow$  partial overlap;  $y(t) = \text{linear with neg. slope}$

$4 < t \Rightarrow$  no overlap;  $y(t) = 0$



$\leftarrow z(t)$

Problem 3. (cont'd.)

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(b) [15 pts.] Find the impulse response of the discrete-time system described by the difference equation:

$$y[n] + 2y[n-1] = x[n-1].$$

Let  $x[n] = \delta[n]$

$\Rightarrow y[n] = h[n]$  with  $h[-1] = 0$  at rest before application of input.

$$h[n] + 2h[n-1] = \delta[n-1]$$

$n=0$      $h[0] + 2h[-1] = \delta[-1] \Rightarrow h[0] = 0$

$n=1$      $h[1] + 2h[0] = \delta[0] \Rightarrow h[1] = 1$

$n=2$      $h[2] + 2h[1] = 0 \Rightarrow h[2] = -2$

$n=3$      $h[3] + 2h[2] = 0 \Rightarrow h[3] = 4$

$n=4$      $h[4] + 2h[3] = 0 \Rightarrow h[4] = -8$

$$\therefore h[n] = \begin{cases} 0 & n = 0 \\ (-2)^{n-1} & n \geq 1 \end{cases}$$

or  $h[n] = (-2)^{n-1} u[n-1]$