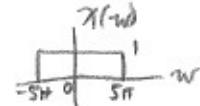


1. Consider the following CT signals:

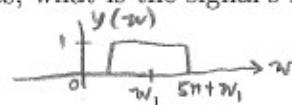
$$\begin{aligned}x(t) &= \frac{\sin(5\pi t)}{\pi t}, \\y(t) &= x(t)e^{j\omega_1 t}, \\z(t) &= x(t)\cos(\omega_2 t) = x(t)\left(\frac{1}{2}e^{j\omega_2 t} + \frac{1}{2}e^{-j\omega_2 t}\right)\end{aligned}$$

(5 pts) a) Is the signal $x(t)$ band limited? (Answer yes/no and justify.)
If you answered yes, what is the signal's Nyquist rate?

yes, $\mathcal{X}(w) \stackrel{\text{By 19}}{=} u(w+5\pi) - u(w-5\pi) =$ 

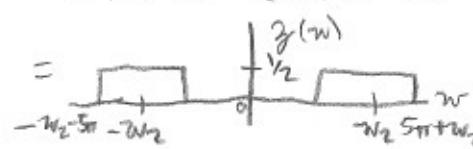
$$\rightarrow w_m = 5\pi \rightarrow w_N = 2w_m = 10\pi$$

(5 pts) b) Assuming that $\omega_1 > 0$, is the signal $y(t)$ band limited? (Answer yes/no and justify.) If you answered yes, what is the signal's Nyquist rate?

yes, $y(w) \stackrel{\text{By 12}}{=} \mathcal{X}(w - \omega_1) =$ 

$$\rightarrow w_m = 5\pi + \omega_1 \rightarrow w_N = 2w_m = 10\pi + 2\omega_1$$

(5 pts) c) Assuming that $\omega_2 > 0$, is the signal $z(t)$ band limited? (Answer yes/no and justify.) If you answered yes, what is the signal's Nyquist rate?

$$\begin{aligned}\text{yes, } y(w) &= \frac{1}{2\pi} \mathcal{X}(w) * (2\pi(\frac{1}{2}\delta(w-\omega_2) + \frac{1}{2}\delta(w+\omega_2))) \\&= \frac{1}{2} \mathcal{X}(w) * (\delta(w-\omega_2) + \delta(w+\omega_2)) \\&= \frac{1}{2} \left[u(w-\omega_2) - u(w+\omega_2) \right]\end{aligned}$$


$$\rightarrow w_m = 5\pi + \omega_2 \rightarrow w_N = 2w_m = 10\pi + 2\omega_2$$

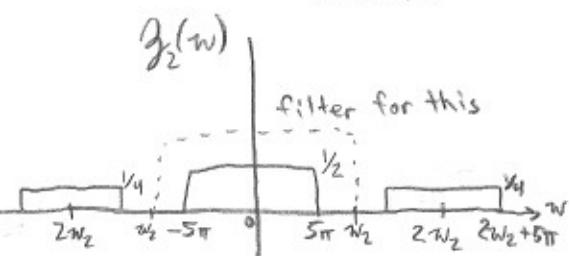
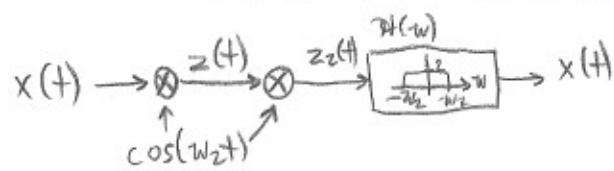
(10 pts) d) Can one recover $x(t)$ from $y(t)$? (Answer yes/no. If you answer yes, explain how. If you answered no, explain why not.)

yes, multiply $y(t)$ by $e^{-jw_1 t}$, By 12, this will shift $y(w)$ back to $X(w)$,

$$\mathcal{F}(y(t)e^{-jw_1 t}) = X(w - w_1 + w_1) = X(w)$$

(15 pts) e) Can one recover $x(t)$ from $z(t)$? (Answer yes/no. If you answer yes, explain how. If you answered no, explain why not.)

yes, multiply $z(t)$ by $\cos(w_2 t)$ and then low pass filter with a filter of gain 2 and $w_{c0} = w_2$



(15 pts) f) Impulse-train sampling is used to obtain

$$x_p[n] = \sum_{k=-\infty}^{\infty} x(n)\delta(n - kN).$$

If the sampling period is $N = \frac{2}{11}$, will aliasing occur? Justify your answer.

$$\omega_s = \frac{2\pi}{N} = \frac{2\pi}{2/11} = 2\pi \cdot \frac{11}{2} = 11\pi$$

and $\omega_N = 10\pi$ (from a.)

$\Rightarrow \omega_s > \omega_N \rightarrow$ aliasing will not occur

(15 pts) g) Impulse-train sampling is used to obtain

$$y_p[n] = \sum_{k=-\infty}^{\infty} y(n)\delta(n - kN).$$

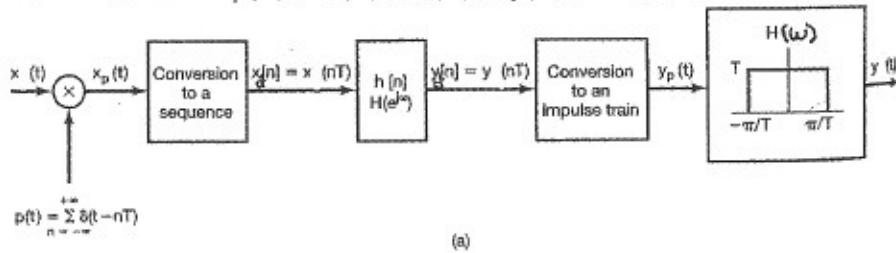
If $\omega_1 = -3\pi$ and the sampling period is $N = \frac{1}{6}$, will aliasing occur? Justify your answer.

$$\omega_s = \frac{2\pi}{N} = \frac{2\pi}{\frac{1}{6}} = 2\pi \cdot 6 = 12\pi$$

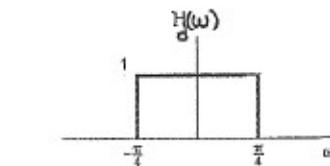
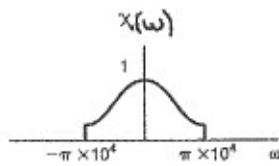
$$\omega_N = 10\pi + 2\omega_1 = 10\pi - 6\pi = 4\pi \quad (\text{from b.})$$

$\Rightarrow \omega_s > \omega_N \Rightarrow$ aliasing will not occur

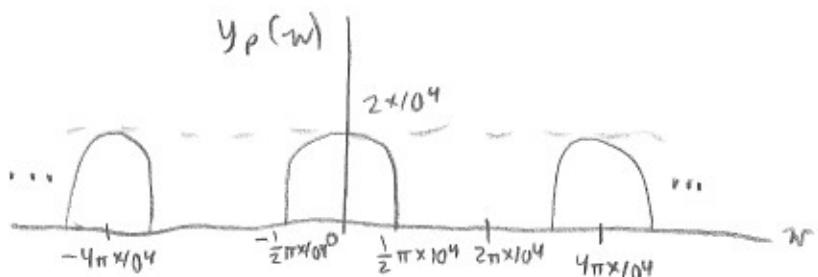
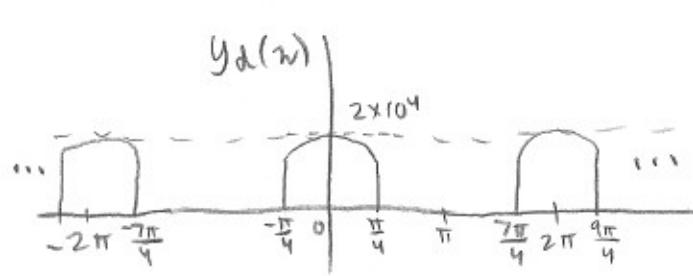
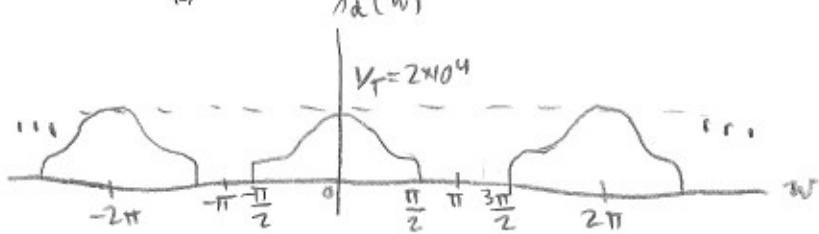
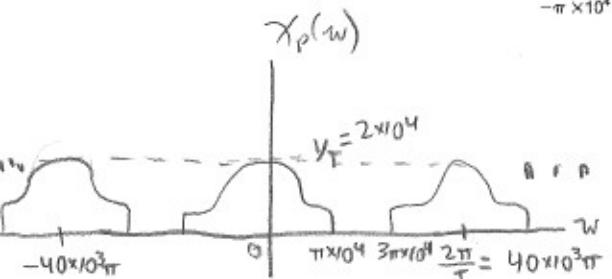
(20 pts) 2. The following figure shows the overall system for filtering a CT signal using a DT filter. If $\mathcal{X}(\omega)$ and $\mathcal{H}_d(\omega)$ are as shown below, with $\frac{1}{T} = 20\text{kHz}$, sketch $\mathcal{X}_p(\omega)$, $\mathcal{X}_d(\omega)$, $\mathcal{Y}_d(\omega)$, $\mathcal{Y}_p(\omega)$, and $\mathcal{Y}(\omega)$.



(a)



(b)



(15 pts) 3. Using the definition of the Laplace transform (i.e. do not simply take the answer from the table), compute the Laplace transform of

$$x(t) = e^{-5t}u(-t)$$

$$\begin{aligned}
 X(s) &= \int_{-\infty}^{\infty} x(t) e^{-st} dt = \int_{-\infty}^{\infty} e^{-st} u(-t) e^{-st} dt \\
 &= \int_{-\infty}^{0} e^{-(s+5)t} dt \\
 &= \begin{cases} \text{divergent}, s = -5 \\ \left[\frac{-1}{s+5} e^{-(s+5)t} \right]_0^{-\infty}, \text{else} \end{cases} = \begin{cases} \text{divergent}, s = -5 \\ \text{divergent}, s + 5 > 0 = s > -5 \\ \left[\frac{-1}{s+5} e^{-(s+5)t} \right]_0^{-\infty}, \text{else} \end{cases} \\
 &= \begin{cases} \text{divergent}, s \geq -5 \\ \frac{-1}{s+5}, \text{else} \end{cases} \\
 &= \frac{-1}{s+5}, \text{ROC: } s < -5
 \end{aligned}$$