

(15 pts) 1. Compute the Fourier transform of the DT signal

$$x[n] = n^2 u[n-2] - n^2 u[n+2]$$

(Express your answer as a linear combination of sine and/or cosine functions.)

$$X(\omega) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n}$$

$$= \sum_{n=-\infty}^{\infty} n^2 e^{-j\omega n} [u[n-2] - u[n+2]]$$

$$= \sum_{n=-2}^{\infty} n^2 e^{-j\omega n} - \sum_{n=-\infty}^{-2} n^2 e^{-j\omega n}$$

$$= (-2)^2 e^{-j\omega(-2)} - (-1)^2 e^{-j\omega(-1)} - (0) - (1)^2 e^{-j\omega(1)} - (2)^2 e^{-j\omega(2)}$$

$$= 4e^{2j\omega} - 4e^{-2j\omega} - 1e^{-j\omega} - 1e^{-j\omega}$$

$$= 8 \cos 2\omega - 2 \cos \omega$$

(15 pts) 2. Show that the Fourier transform of the CT signal $x(t) = \cos(\omega_0 t)$ is $X(\omega) = \pi\delta(\omega + \omega_0) + \pi\delta(\omega - \omega_0)$.

$$X(\omega) = \frac{1}{2\pi} \int_{-\infty}^{\infty} \pi \delta(\omega + \omega_0) e^{j\omega n} + \pi \delta(\omega - \omega_0) e^{j\omega n} d\omega$$

$$= \frac{1}{2\pi} \pi e^{j\omega n} \Big|_{-\omega_0} + \frac{1}{2\pi} \pi e^{j\omega n} \Big|_{\omega_0}$$

$$= \frac{1}{2} e^{j(-\omega_0)n} + \frac{1}{2} e^{j\omega_0 n}$$

$$= \boxed{\cos(\omega_0 t)}$$

(15 pts) 3. Given is a DT signal $x[n] = \frac{1}{g[n]^2}$ where $g[n]$ is a pure imaginary signal and an odd function of n .

a) Bob claims that the Fourier transform of $x[n]$ is $X(\omega) = \frac{j}{\cos \omega}$. Explain why Bob's answer is wrong.

If $x[n]$ is real and even, $X(\omega)$ must be real and even.
However, $\frac{j}{\cos \omega}$ is pure imaginary and even.

b) Alice says that the Fourier transform of $x[n]$ is $X(\omega) = \frac{1}{\sin \omega}$. Could Alice be right? Explain.

No, If $x[n]$ is real and even, $X(\omega)$ must be real and even.
However, $\frac{1}{\sin \omega}$ is real and odd.

c) Devin says that the Fourier transform of $x[n]$ is $X(\omega) = \frac{1}{\omega^2}$. Could Devin be right? Explain.

No. Fourier transforms in DT must be periodic with period 2π .

4. A discrete-time LTI system has frequency response

$$H(e^{j\omega}) = \frac{2}{1 - \frac{3}{4}e^{-j\omega} + \frac{1}{8}e^{-2j\omega}}$$

(15 pts) a) Derive a difference equation relating the input and the output of this system. (Use the properties of the Fourier transform listed in the table to justify your answer.)

$$Y(e^{j\omega}) = X(e^{j\omega}) \cdot H(e^{j\omega}) = \frac{2X(e^{j\omega})}{1 - \frac{3}{4}e^{-j\omega} + \frac{1}{8}e^{-2j\omega}}$$

$$Y(e^{j\omega}) - \frac{3}{4}Y(e^{j\omega})e^{-j\omega} + \frac{1}{8}Y(e^{j\omega})e^{-2j\omega} = 2X(e^{j\omega})$$

by linearity and time shifting:

$$y[n] - \frac{3}{4}y[n-1] + \frac{1}{8}y[n-2] = 2x[n]$$

(10 pts) b) What is the Fourier transform of the output when the input is $x[n] = (\frac{1}{4})^n u[n]$?

$$X(\omega) = \frac{1}{1 - (\frac{1}{4})e^{-j\omega}}$$

$$Y(\omega) = X(\omega) \cdot H(\omega) = \frac{2}{[1 - (\frac{1}{4})e^{-j\omega}][1 - \frac{3}{4}e^{-j\omega} + \frac{1}{8}e^{-2j\omega}]}$$

(15 pts) b) Find the unit impulse response of this system.

$$H(e^{j\omega}) = \frac{2}{\left(\frac{1}{8}\right)(e^{-j\omega})^2 + \left(\frac{3}{4}\right)(e^{-j\omega}) + 1} = \frac{16}{(e^{-j\omega})^2 - 6(e^{-j\omega}) + 8} = \frac{16}{(e^{-j\omega} - 4)(e^{-j\omega} - 2)}$$

(by partial fraction expansion)

$$= \frac{8}{e^{-j\omega} - 4} + \frac{-8}{e^{-j\omega} - 2} = \frac{-2}{1 - \frac{1}{4}e^{-j\omega}} + \frac{4}{1 - \frac{1}{2}e^{-j\omega}}$$

$$F^{-1}(H(e^{j\omega})) = h[n] = -2 F^{-1}\left(\frac{1}{1 - \frac{1}{4}e^{-j\omega}}\right) + 4 F^{-1}\left(\frac{1}{1 - \frac{1}{2}e^{-j\omega}}\right)$$

$$h[n] = \left[-2\left(\frac{1}{4}\right)^n + 4\left(\frac{1}{2}\right)^n \right] u[n]$$

(20 pts) 5. Use the definition of the Fourier transform (*not* the properties listed in the table) to prove the following Fourier transform property.

$$x(at+b) \xrightarrow{F} \frac{e^{j\omega \frac{b}{a}}}{-a} \mathcal{X}\left(\frac{\omega}{a}\right) \text{ for any } a, b \text{ real numbers with } a < 0.$$

$$X(\omega) = \int_{-\infty}^{\infty} x(at+b) e^{-j\omega t} dt$$

(let $z = at+b$ and $dz = a dt$)

$$= \frac{1}{a} \int_{\infty}^{-\infty} x(z) e^{-j\omega \left(\frac{z-b}{a}\right)} dz$$

$$a(-\infty)+b \Rightarrow \infty \quad \text{b/c } a < 0$$

$$a(\infty)+b \Rightarrow -\infty$$

$$-e^{j\omega \frac{b}{a}} \left(\frac{1}{a}\right) \int_{-\infty}^{\infty} x(z) e^{-j\omega \left(\frac{z}{a}\right)} dz =$$

$$\boxed{\frac{e^{j\omega \frac{b}{a}}}{-a} \mathcal{X}\left(\frac{\omega}{a}\right)}$$

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