

1. Determine whether each of the following signals is band limited. (Answer yes/no and justify.) If they are band limited, specify their Nyquist rate.

(5 pts) a) $x_1(t) = \frac{\sin(2\pi t)}{t} \xrightarrow{\mathcal{F}} \pi \left(\frac{\sin(2\pi t)}{\pi t} \right) \xrightarrow{\mathcal{F}} \pi u(\omega + 2\pi) - \pi u(\omega - 2\pi)$

$NQ = 2\omega_{max}, \omega_{max} = 2\pi \Rightarrow NQ = 4\pi$



(5 pts) b) $x_2(t) = 3(u(t+2) - u(t-2)) \xrightarrow{\mathcal{F}} 3 \left(\frac{2 \sin(\omega 2)}{\omega} \right)$

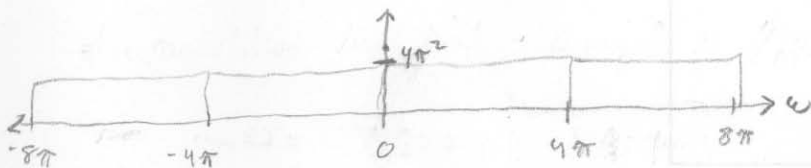
There is no value ω_m for which $\mathcal{F}(x_2(t)) = 0$ when $\omega > \omega_m$

$\therefore x_2(t)$ is NOT BAND LIMITED



(5 pts) c) $x_3(t) = \frac{\sin(2\pi t)}{t} \sum_{n=-\infty}^{\infty} \delta(t - \frac{n}{2})$

$\xrightarrow{\mathcal{F}} = \pi(u(\omega + 2\pi) - u(\omega - 2\pi)) * \frac{2\pi}{1/2} \sum_{k=-\infty}^{\infty} \delta(\omega - \frac{2\pi k}{1/2}), \text{ by (14)(19)(22)}$
 $= 4\pi^2 \left((u(\omega + 2\pi) - u(\omega - 2\pi)) * \sum_{k=-\infty}^{\infty} \delta(\omega - 4\pi k) \right)$



$\therefore \omega_m$ does not exist. NOT BAND LIMITED

(5 pts) d) $x_4(t) = \left(\frac{\sin(2\pi t)}{t}\right)^2 \xrightarrow{\mathcal{F}} \pi(u(\omega+2\pi) - u(\omega-2\pi))$, but since it is squared,

the frequency is doubled and $\omega_{max} = 2 \cdot 2\pi = 4\pi$

$$NQ = 2 \cdot \omega_m = 2 \cdot 4\pi = \boxed{8\pi = NQ}$$

(5 pts) d) $x_5(t) = \frac{\sin(2\pi t)}{t} \cos(5\pi t) \xrightarrow{\mathcal{F}} \left(\frac{e^{j5\pi t} + e^{-j5\pi t}}{2}\right)$

$\xrightarrow{\mathcal{F}} \pi(u(\omega+2\pi) - u(\omega-2\pi)) * \frac{1}{2}(2\pi\delta(\omega-5\pi) + 2\pi\delta(\omega+5\pi))$, by (13)(10)(14)
(17)(17)

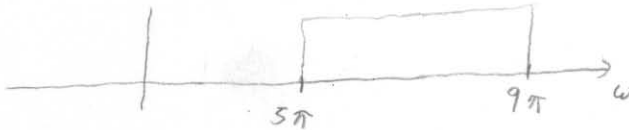


$\therefore |X_5(\omega)| = 0$ when $|\omega| > 9\pi$ $NQ = 2\omega_m$ $\omega_m = 9\pi$ $\boxed{NQ = 18\pi}$

(5 pts) e) $x_6(t) = \frac{\sin(2\pi t)}{t} e^{j5\pi t}$

$\xrightarrow{\mathcal{F}} \pi(u(\omega+2\pi) - u(\omega-2\pi)) * 2\pi\delta(\omega-5\pi)$ by (14)(17)(19)

$= 2\pi^2(u(\omega+2\pi) - u(\omega-2\pi)) * \delta(\omega-5\pi)$



$\omega_m = 9\pi$

$$\boxed{NQ = 18\pi}$$

(15 pts) 2. Using the definition of the z-transform (i.e. do not simply take the answer from the table), obtain the z-transform (with its ROC) of

$$x[n] = 3^n u[-n - 1]$$

$$X(z) = \sum_{n=-\infty}^{\infty} x[n] z^{-n}$$

$$= \sum_{n=-\infty}^{\infty} 3^n u[-n-1] z^{-n}$$

$$= \sum_{n=-\infty}^{-1} 3^n z^{-n}$$

let $k = -n$

$$= \sum_{k=\infty}^1 3^{-k} z^k$$

$$= \sum_{k=1}^{\infty} \left(\frac{1}{3}\right)^k z^k$$

let $m = k-1 \Rightarrow k = m+1$

$$= \sum_{m=0}^{\infty} \left(\frac{z}{3}\right)^{m+1} = \frac{z}{3} \sum_{m=0}^{\infty} \left(\frac{z}{3}\right)^m$$

$$= \frac{z}{3} \left(\frac{1}{1 - \frac{z}{3}} \right) \quad \text{ROC: } -3 < z < 3$$

(20 pts) 3. The Laplace transform of the unit impulse response of a system is

$$H(s) = \frac{1}{s+2}, \operatorname{Re}(s) > -2.$$

Determine the response $y(t)$ of the system when the input is

$$x(t) = \begin{cases} e^{-3t} & t > 0 \\ 0 & t = 0 \\ -e^{3t} & t < 0 \end{cases}$$

$$x(t) = e^{-3t} u(t) - e^{3t} u(-t) = e^{-3t} (u(t) - u(-t))$$

$$\mathcal{L}(x(t)) = \begin{cases} \frac{1}{s+3}, & \operatorname{Re}(s) > -3 \\ \frac{1}{s+3}, & \operatorname{Re}(s) < -3 \\ \text{undefined}, & \operatorname{Re}(s) = -3 \end{cases} \Rightarrow \begin{cases} \text{undefined}, & \operatorname{Re}(s) = -3 \\ \frac{1}{s+3}, & \text{else} \end{cases}$$

$$Y(s) = H(s) X(s)$$

$$Y(s) = \frac{1}{s+2} \cdot \frac{1}{s+3} = \frac{A}{s+2} + \frac{B}{s+3} = \frac{1}{s+2} - \frac{1}{s+3}$$

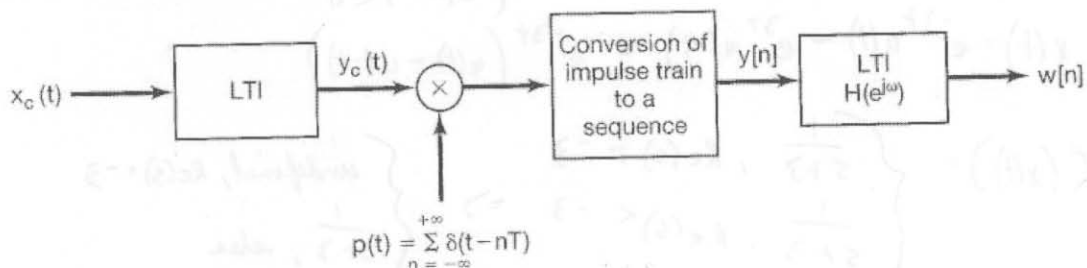
$$\begin{array}{l} \downarrow \mathcal{L}^{-1} \\ \operatorname{Re}(s) > -2 \end{array} \quad \begin{array}{l} \downarrow \mathcal{L}^{-1} \\ \operatorname{Re}(s) \neq -3 \end{array}$$

$$y(t) = e^{-2t} u(t) - e^{-3t} u(t)$$

4. The block diagram below shows a system consisting of a continuous-time LTI system followed by a sampler, conversion to a sequence, and an LTI discrete-time system. The continuous-time LTI system is causal and satisfies the linear, constant-coefficient differential equation

$$\frac{dy_c(t)}{dt} + y_c(t) = x_c(t).$$

The input $x_c(t)$ is a unit impulse $\delta(t)$.



(10 pts) a) Determine the input $y_c(t)$.

$$s Y_c(s) + Y_c(s) = X(s), \text{ by (40) (46)}$$

$$Y_c(s) = \frac{X(s)}{s+1}$$

$$X(s) = 1 \text{ by (54)}$$

$$Y_c(s) = \frac{1}{s+1}$$

\downarrow \mathcal{L}^{-1} - causal systems are right sided $\therefore \mathcal{L.T.}$ is also right sided.

$$y_c(t) = e^{-t} u(t), \text{ by (52)}$$

(Problem 7 continues on the next page.)

(15 pts) b) Determine the frequency response $H(e^{j\omega})$ and the unit impulse response $h[n]$ such that $w[n] = \delta[n]$.

$$H(\omega) = \frac{y(\omega)}{x(\omega)} = \frac{y(\omega)}{1}$$

to cancel that
out so that $w(\omega) = 1$

$$H(\omega) = \frac{x(\omega)}{y(\omega)}$$

$$H(\omega) = \frac{1}{1 - e^{-T}e^{-j\omega}}$$

$$H(\omega) = 1 - e^{-T}e^{-j\omega}$$

$$\mathcal{F}^{-1} \rightarrow \delta[n] - e^{-T}\delta[n-1] \quad \text{by (3b) (2c)}$$

$$h[n] = \delta[n] - e^{-T}\delta[n-1]$$

$$y_c(t) = e^{-t} u(t)$$

$$y_c[nT] = e^{-nT} u[nT]$$

$$= (e^{-T})^n u[nT]$$

$$y[n] = (e^{-T})^n u[n]$$

$$\mathcal{F} \rightarrow \frac{1}{1 - (e^{-T})e^{-j\omega}} \quad \text{when } |e^{-T}| < 1 \quad \text{by (3b)}$$

$$y(\omega) = \frac{1}{1 - e^{-T}e^{-j\omega}}$$