

Problem 1  $x(t) \leftrightarrow X(j\omega)$

(a)  $x^*(t) \leftrightarrow X^*(-j\omega)$  directly from Table. If we desire to prove directly

$$\begin{aligned}
 x(t) &= \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{+j\omega t} d\omega \\
 x^*(t) &= \frac{1}{2\pi} \int_{-\infty}^{\infty} X^*(j\omega) e^{-j\omega t} d\omega \quad \left. \begin{array}{l} \text{conjugate} \\ \text{c.o.v.} \\ \omega \mapsto -\omega \\ d\omega \mapsto -d\omega \end{array} \right\} \\
 &= \frac{1}{2\pi} \int_{-\infty}^{\infty} X^*(-j\omega) e^{+j\omega t} (-d\omega) \quad \left. \begin{array}{l} \text{flip limits} \end{array} \right\} \\
 &= \frac{1}{2\pi} \int_{-\infty}^{\infty} X^*(-j\omega) e^{+j\omega t} d\omega
 \end{aligned}$$

(b)  $\text{Re}\{x(t)\}$

$$\begin{aligned}
 \text{Note } \text{Re}\{x(t)\} &= \frac{x(t) + x^*(t)}{2} \\
 &\quad \updownarrow \\
 \text{Fourier transform} &= \frac{X(j\omega) + X^*(-j\omega)}{2}
 \end{aligned}$$

(c)  $x(t-2) + x^*(-t-2)$

$$\begin{aligned}
 x(t-2) &\leftrightarrow e^{-j2\omega} X(j\omega) \quad \text{from Table} \\
 x^*(t-2) &\leftrightarrow [e^{-j2\omega} X(j\omega)]^* \Big|_{j\omega \mapsto -j\omega} \\
 &= e^{+j2\omega} X^*(j\omega) \Big|_{j\omega \mapsto -j\omega} \\
 &= e^{-j2\omega} X^*(-j\omega)
 \end{aligned}$$

$$x^*(-t-z) \leftrightarrow \left[ e^{-jz\omega} X^*(-j\omega) \right] \Big|_{j\omega \leftrightarrow -j\omega}$$

$$= e^{+jz\omega} X^*(j\omega) \quad \text{from Table entries}$$

$$\therefore x(t-z) + x^*(-t-z) \leftrightarrow \underbrace{e^{-jz\omega} X(j\omega) + e^{+jz\omega} X^*(j\omega)}$$

Can also write as  $X(j\omega) = |X(j\omega)| e^{j\angle X(j\omega)}$

$$|X(j\omega)| \left\{ e^{j(z\omega - \angle X(j\omega))} + e^{-j(z\omega - \angle X(j\omega))} \right\}$$

$$= 2|X(j\omega)| \cos(z\omega - \angle X(j\omega)).$$

$$(d) e^{j4\pi t} x(t/3)$$

$$x(t/3) \leftrightarrow 3X(j3\omega)$$

$$e^{j4\pi t} x(t/3) \leftrightarrow 3X(j3(\omega - 4\pi)).$$

Problem 2

$$BW = \underbrace{\frac{1}{|H(j\omega)|^2}}_{\substack{\text{this from} \\ \text{transform} \\ \text{definition}}} \underbrace{\frac{1}{2\pi} \int_{-\infty}^{\infty} |H(j\omega)|^2 d\omega}_{\text{this from Parseval}}$$

$$H(j\omega) = \int_{-\infty}^{\infty} h(t) e^{-j\omega t} dt \Rightarrow |H(j\omega)| = \left| \int_{-\infty}^{\infty} h(t) dt \right|$$

$$= 2 - 1 + 1 - 1$$

$$= 1$$

$$\frac{1}{2\pi} \int_{-\infty}^{\infty} |H(j\omega)|^2 d\omega = \int_{-\infty}^{\infty} |h(t)|^2 dt$$

$$= 4 \cdot 1 + 1 \cdot 1 + 1 \cdot 1 + 1 \cdot 1$$

$$= 7$$

$$\therefore BW = \frac{7}{1} = 7.$$

### Problem 3

$$(a) \quad z_I(t) = x_I(t) \cos \omega_c t$$



$$Z_I(j\omega) = \frac{1}{2\pi} X_I(j\omega) * \pi [\delta(\omega - \omega_c) + \delta(\omega + \omega_c)]$$

$$= \frac{1}{2} X_I(j(\omega - \omega_c)) + \frac{1}{2} X_I(j(\omega + \omega_c))$$

$$z_Q(t) = x_Q(t) \sin \omega_c t$$



$$Z_Q(j\omega) = \frac{1}{2\pi} X_Q(j\omega) * \frac{\pi}{j} [\delta(\omega - \omega_c) - \delta(\omega + \omega_c)]$$

$$= -\frac{j}{2} X_Q(j(\omega - \omega_c)) + \frac{j}{2} X_Q(j(\omega + \omega_c))$$

Since  $X_I(j\omega)$  and  $X_Q(j\omega)$  are assumed to be real-valued see that

$Z_I(j\omega)$  is pure real

$Z_Q(j\omega)$  is pure imaginary

$$Z(j\omega) = Z_I(j\omega) + Z_Q(j\omega)$$

$$= \left[ \frac{X_I(j(\omega - \omega_c)) + X_I(j(\omega + \omega_c))}{2} \right]$$

$\text{Re}\{Z(j\omega)\}$

$$+ j \left[ \frac{-X_Q(j(\omega - \omega_c)) + X_Q(j(\omega + \omega_c))}{2} \right]$$

$\text{Im}\{Z(j\omega)\}$

(b) Since  $h_c(t) = \delta(t)$  here also have  $H_c(j\omega) = 1$   
Therefore  $Z_c(t) = Z(t)$ .

$$S_I(t) = Z(t) 2 \cos \omega_c t$$

↕

$$S_I(j\omega) = \frac{1}{2\pi} Z(j\omega) * 2\pi [\delta(\omega - \omega_c) + \delta(\omega + \omega_c)]$$

$$= Z(j(\omega - \omega_c)) + Z(j(\omega + \omega_c)) \leftarrow \text{this is enough to plot using part (a).}$$

(if want to write in terms of  $X_I$  and  $X_Q$ )

$$= \left[ \frac{1}{2} X_I(j(\omega - 2\omega_c)) + X_I(j\omega) + \frac{1}{2} X_I(j(\omega + 2\omega_c)) \right]$$

$$+ j \left[ -\frac{1}{2} X_Q(j(\omega - 2\omega_c)) + \frac{1}{2} X_Q(j(\omega + 2\omega_c)) \right]$$

$$S_Q(t) = Z(t) 2 \sin \omega_c t$$

↕

$$S_Q(j\omega) = \frac{1}{2\pi} Z(j\omega) * \frac{2\pi}{j} [\delta(\omega - \omega_c) - \delta(\omega + \omega_c)]$$

$$= -j Z(j(\omega - \omega_c)) + j Z(j(\omega + \omega_c)) \leftarrow \text{enough to plot using (a)}$$

(in terms of  $X_I$  and  $X_Q$ )

$$= \left[ -\frac{1}{2} X_Q(j(\omega - 2\omega_c)) + X_Q(j\omega) - \frac{1}{2} X_Q(j(\omega + 2\omega_c)) \right]$$

$$+ j \left[ -\frac{1}{2} X_I(j(\omega - 2\omega_c)) + \frac{1}{2} X_I(j(\omega + 2\omega_c)) \right]$$

(c) The LP filters will select on the frequencies  $|\omega| \leq \omega_M$  in both  $S_I(j\omega)$  and  $S_Q(j\omega)$ .

Note from picture that the imaginary parts are zero for  $|\omega| \leq \omega_M$ .

$$\therefore Y_I(j\omega) = H_{LP}(j\omega) S_I(j\omega) = X_I(j\omega)$$

$$Y_Q(j\omega) = H_{LP}(j\omega) S_Q(j\omega) = X_Q(j\omega).$$

(d) Now assume a general channel filter  $H_c(j\omega)$ .

Then  $Z_c(j\omega) = H_c(j\omega) Z(j\omega)$  and looking only at the in-phase branch

$$Y_I(j\omega) = H_{LP}(j\omega) S_I(j\omega)$$

$$= H_{LP}(j\omega) \left[ Z_c(j(\omega - \omega_c)) + Z_c(j(\omega + \omega_c)) \right]$$

$$= H_{LP}(j\omega) \left[ H_c(j(\omega - \omega_c)) Z(j(\omega - \omega_c)) + H_c(j(\omega + \omega_c)) Z(j(\omega + \omega_c)) \right]$$

$$\left[ \right] = H_c(j(\omega - \omega_c)) \left\{ \frac{1}{2} X_I(j(\omega - 2\omega_c)) + \frac{1}{2} X_I(j\omega) - \frac{j}{2} X_Q(j(\omega - 2\omega_c)) + \frac{j}{2} X_Q(j\omega) \right\}$$

$$+ H_c(j(\omega + \omega_c)) \left\{ \frac{1}{2} X_I(j\omega) + \frac{1}{2} X_I(j(\omega + 2\omega_c)) - \frac{j}{2} X_Q(j\omega) + \frac{j}{2} X_Q(j(\omega + 2\omega_c)) \right\}$$

Now apply  $H_{LP}(j\omega)$  to  $[\ ]$  above will remove the spectral terms centered at  $\pm 2\omega_c$ :

$$\begin{aligned}
 & H_{LP}(j\omega) [\ ] \\
 &= H_{LP}(j\omega) H_c(j(\omega - \omega_c)) \left\{ \frac{1}{2} X_I(j\omega) + \frac{j}{2} X_Q(j\omega) \right\} \\
 &+ H_{LP}(j\omega) H_c(j(\omega + \omega_c)) \left\{ \frac{1}{2} X_I(j\omega) - \frac{j}{2} X_Q(j\omega) \right\} \\
 &= \frac{1}{2} X_I(j\omega) H_{LP}(j\omega) [H_c(j(\omega - \omega_c)) + H_c(j(\omega + \omega_c))] \\
 &+ \frac{j}{2} X_Q(j\omega) H_{LP}(j\omega) [H_c(j(\omega - \omega_c)) - H_c(j(\omega + \omega_c))]
 \end{aligned}$$

This must be zero to have no cross talk. Since channel filter has a real-valued imp. response  $H_c(j\omega) = H_c^*(-j\omega)$

$$\Rightarrow H_c(j(\omega - \omega_c)) = H_c^*(j(\omega_c - \omega))$$

$\therefore$  Suff. cond. for no cross talk

$$H_{LP}(j\omega) [H_c^*(j(\omega_c - \omega)) - H_c(j(\omega_c + \omega))] = 0 \quad \forall \omega.$$

