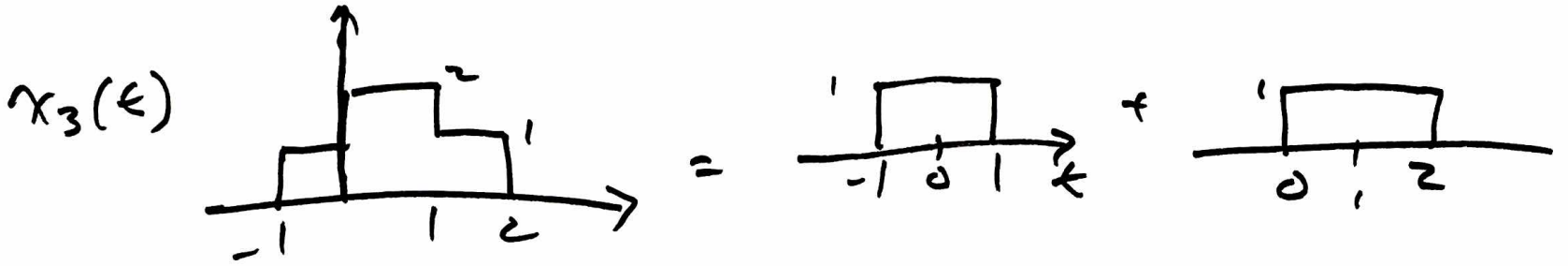
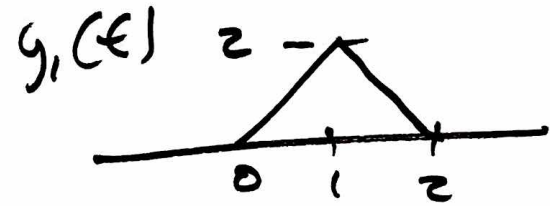
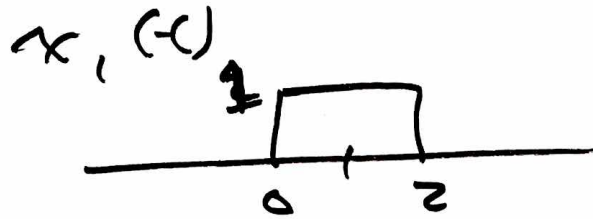


1.31



$$x_3(t) = \sum_{k=0}^{\infty} a_k x_1(t - \tau_k)$$

$$\begin{aligned} S\{x_3(t)\} &= S\left\{\sum_{k=0}^{\infty} a_k x_1(t - \tau_k)\right\} \\ &= \sum_{k=0}^{\infty} a_k \underbrace{S\{x_1(t - \tau_k)\}}_{y_1(t - \tau_k)} \end{aligned}$$

$$\begin{aligned} y(t) &= x(t) * h(t) \\ y\left(\frac{t}{3}\right) &= x\left(\frac{t}{3}\right) * h\left(\frac{t}{3}\right) \\ &\neq x\left(\frac{t}{3}\right) * h(t) \end{aligned}$$

Time shift in  $\rightarrow$  time shift out  
 amplitude scaled in  $\rightarrow$  amplitude scaled out  
 time scaled in  $\rightarrow$  time scaled out

①

~~DT~~ DT systems described by constant coefficient difference equations

- Use difference eqn to describe causal LTI systems

Ex

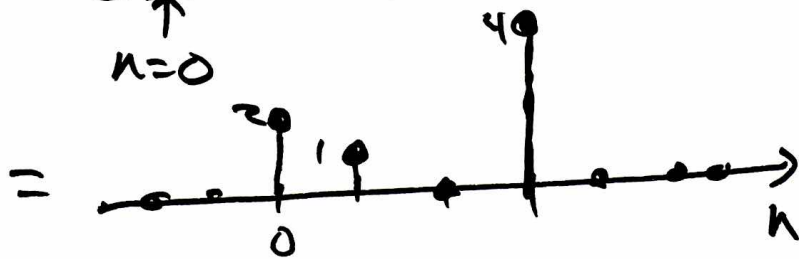
$$y[n] = 2x[n] + x[n-1] + 4x[n-3]$$

$y[n]$  depends only on  $x[n-k]$

$$x[n] = \delta[n] \Rightarrow y[n] = h[n]$$

$$h[n] = 2\delta[n] + \delta[n-1] + 4\delta[n-3] \quad \checkmark$$

$$= \{ \underset{\substack{\uparrow \\ n=0}}{2}, 1, 0, 4 \}$$



Ex

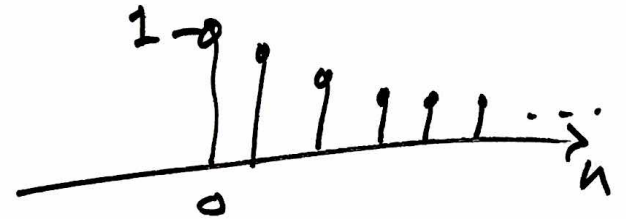
$$y[n] = 0.5y[n-1] + x[n]$$

$$x[n] = \delta[n] \Rightarrow y[n] = h[n]$$

$$h[n] = 0.5h[n-1] + \delta[n]$$

n	h[n]	h[n-1]	$\delta[n]$
0	1	0	1
1	0.5	1	0
2	$(0.5)^2$	0.5	0
3	$(0.5)^3$	$(0.5)^2$	0
4	$(0.5)^4$	$(0.5)^3$	0
	⋮	⋮	⋮

$$h[n] = (0.5)^n u[n]$$



$$\sum_{k=0}^N a_k y[n-k] = \sum_{k=0}^M b_k x[n-k]$$

$$a_0 y[n] + \sum_{k=1}^N a_k y[n-k] = \sum_{k=0}^M b_k x[n-k]$$

$$y[n] = \frac{1}{a_0} \left[ \sum_{k=0}^M b_k x[n-k] - \sum_{k=1}^N a_k y[n-k] \right]$$

General form

$$N=0 \quad y[n] = \frac{1}{a_0} \sum_{k=0}^M b_k x[n-k]$$

$$x[n] = \delta[n] \Rightarrow y[n] = h[n]$$

$$h[n] = \frac{1}{a_0} \sum_{k=0}^M b_k \delta[n-k]$$

$$= \frac{1}{a_0} \cdot \underbrace{\{b_0; b_1, b_2, \dots, b_M\}}_{n=0}$$

Finite impulse response system/filter  
FIR

- Recursive equations can lead to an infinite impulse response (IIR) system/filter.

OW 2.5

$$x[n] = \begin{cases} 1 & 0 \leq n \leq 9 \\ 0 & \text{else} \end{cases}$$

$$h[n] = \begin{cases} 1 & 0 \leq n \leq N \\ 0 & \text{else} \end{cases}$$

$$y[4] = 5, \quad y[14] = 0$$

In general,  $\text{length}(x[n] * h[n]) \leq \text{length}(x[n]) + \text{length}(h[n]) - 1$  should be =

$$10 + N + 1 - 1 \leq 14$$

not given  $y[13]$

$$N \leq 4$$

$$y[4] = \sum_{k=0}^4 h[4-k] x[k] = 5$$

$$h[4], h[3], h[2], \dots, h[0] = 1$$

$$N \geq 4$$

$$N = 4$$

Where did this come from?  
The length of the output must be  $10 + (N+1) - 1$ . Given that  $y[14] = 0$ , the length of the output must be less than or equal to 14.

Given  $\text{length}(x[n])$  and  $\text{length}(h[n])$  as  $N$  and  $M$ , there are exactly  $N+M-1$  points of overlap between the sequences.

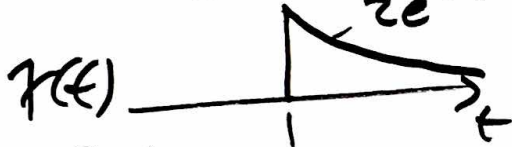
2.46)  $x(t) = 2e^{-3t} u(t-1)$

$x(t) \xrightarrow{s} y(t)$

$\frac{d}{dt} x(t) \xrightarrow{s} -3y(t) + e^{-2t} u(t)$

Find  $h(t)$ ?

$y'(t) = \int \{x'(t)\} = x'(t) * h(t)$



$x'(t) = 2e^{-3} \delta(t-1) + 2(-3)e^{-3t} u(t-1)$   
 $= 2e^{-3} \delta(t-1) - 6e^{-2t} u(t-1)$

$(f \cdot g)' = f'g + f \cdot g'$

$x'(t) = \frac{d}{dt} (2e^{-3t}) u(t-1) + 2e^{-3t} \frac{d}{dt} (u(t-1))$   
 $= -6e^{-3t} u(t-1) + \underbrace{2e^{-3t} \delta(t-1)}_{\text{sampling property}}$   
 $= -6e^{-3t} u(t-1) + 2e^{-3} \delta(t-1)$

$$\cdot (\pi'(t) * h(t)) = \underbrace{(-6e^{-3t} u(t-1) + 2e^{-3} \delta(t-1))}_{-3\pi(t)} * h(t)$$

$$= -3\pi(t) * h(t) + 2e^{-3} \delta(t-1) * h(t)$$

$$= -3y(t) + 2e^{-3} h(t-1) \leftarrow$$

$$= -3y(t) + e^{-2t} u(t) \leftarrow \text{given information}$$

$$2e^{-3} h(t-1) = e^{-2t} u(t)$$

$$h(t-1) = \frac{1}{2} e^{-2t} e^3 u(t)$$

$$= \frac{1}{2} e^3 e^{-2t} u(t)$$

$$h(t) = \frac{1}{2} e^3 e^{-2(t+1)} u(t+1)$$

Takeaway

$$y'(t) = x'(t) * h(t)$$

example

$$s(t) = u(t) * h(t)$$

$$h(t) = \frac{d}{dt} s(t)$$

$$\frac{d}{dt} u(t-\tau) = \delta(t-\tau)$$

