

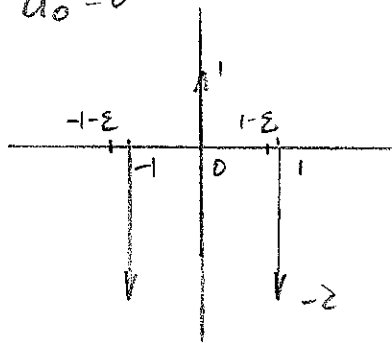
3.22

$$\begin{aligned}
 \underline{a)} \quad a_k &= \frac{1}{T} \int_T x(t) e^{-jk\omega_0 t} dt \\
 &= \frac{1}{2} \int_{-1}^1 t e^{-jk\omega_0 t} dt \\
 &= \frac{1}{2} \left\{ \frac{1}{-jk\omega_0} e^{-jk\omega_0 t} \cdot t \Big|_{-1}^1 - \int_{-1}^1 e^{-jk\omega_0 t} \cdot dt \right\} \\
 &= \frac{1}{2} \left\{ \frac{1}{-jk\omega_0} [e^{-jk\omega_0} + e^{jk\omega_0}] - \frac{1}{-jk\omega_0} e^{-jk\omega_0 t} \Big|_{-1}^1 \right\} \\
 &= \frac{1}{-2jk\omega_0} [2\cos k\omega_0 + 2j \sin k\omega_0] \\
 &= \frac{e^{j(k+\frac{1}{2})\pi}}{k\pi} \quad (*)
 \end{aligned}$$

$$(*) \text{ Note } \omega_0 = \frac{2\pi}{T} = \frac{\pi}{2}$$

$$a_0 = 0$$

d)



$$\begin{aligned}
 a_k &= \frac{1}{2} \int_{-1-\varepsilon}^{1-\varepsilon} [\delta(t) - 2\delta(t-1)] e^{-jk\omega_0 t} dt \\
 &= \frac{1}{2} [e^{-jk\omega_0 \cdot 0} - 2 \cdot e^{-jk\omega_0 (1)}] \\
 &= \frac{1}{2} (1 - 2e^{jk\pi}) \\
 &= \frac{1}{2} [1 - (-1)^k]
 \end{aligned}$$

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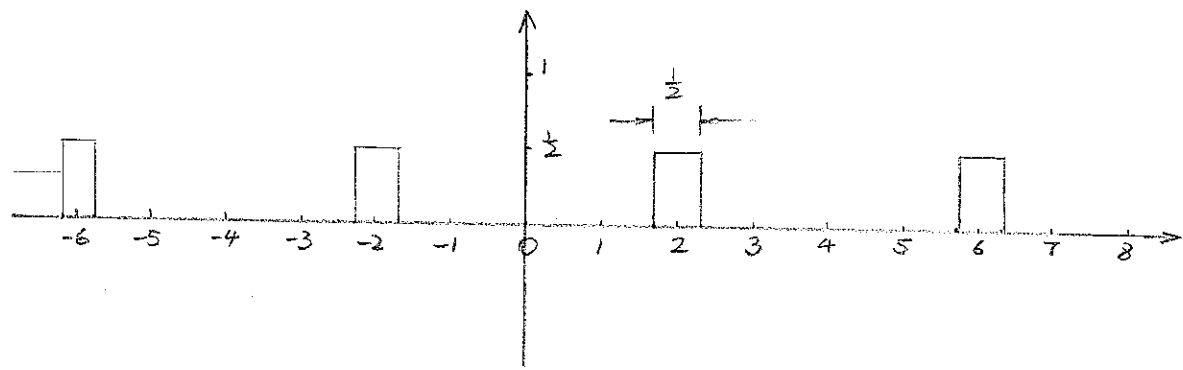
b) It is known that  $C_k = \frac{\sin k\omega_0 T_1}{k\pi}$  is the Fourier Series coefficients of the signal.

$$x_0(t) = \begin{cases} 1, & |t| < T_1 \\ 0, & T_1 < |t| < \frac{T}{2} \end{cases} \quad \text{and } x_0(t+T) = x_0(t).$$

$$\begin{aligned} \text{Now that } a_k &= (-1)^k \frac{\sin k\pi/8}{2k\pi} \\ &= \frac{1}{2} e^{jk\frac{2\pi}{4} \cdot 2} \cdot \frac{\sin k \cdot \frac{2\pi}{4} \cdot \frac{1}{4}}{k\pi} \end{aligned}$$

we know from the time shifting property of Fourier Series, that  $a_k$  must be the Fourier Series coefficients of the signal.

$$x(t) = \begin{cases} \frac{1}{2}, & \frac{7}{4} < t < \frac{9}{4} \\ 0, & -\frac{7}{4} < t < \frac{7}{4} \end{cases} \quad \text{and } x(t+4) = x(t).$$



d)

$$x(t) = \sum_{k=-\infty}^{\infty} a_k e^{jk\omega_0 t}$$

$$= \sum_{n=-\infty}^{\infty} [e^{j2n\omega_0 t} + 2e^{j(2n+1)\omega_0 t}]$$

$$= \sum_{n=-\infty}^{\infty} [e^{j2n \cdot \frac{2\pi}{4} t} + 2e^{j(2n+1) \cdot \frac{2\pi}{4} t}]$$

$$= \sum_{n=-\infty}^{\infty} [e^{jn\pi t} + 2e^{j \cdot \frac{\pi}{2} t} \cdot e^{jn\pi t}]$$

$$= (1 + 2e^{j \cdot \frac{\pi}{2} t}) \sum_{n=-\infty}^{\infty} e^{jn\pi t}$$

$$= (1 + 2e^{j \cdot \frac{\pi}{2} t}) \sum_{k=-\infty}^{\infty} \delta(t - 2k)$$

## ECE 301 Signals and Systems Homework # 5 Solution

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$$\begin{aligned} \text{a) } x(t)y(t) &= \sum_{k=-\infty}^{\infty} a_k e^{jk\omega_0 t} \cdot \sum_{k=-\infty}^{\infty} b_k e^{jk\omega_0 t} \\ &= \sum_{n=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} a_n b_m e^{j(n+m)\omega_0 t} \end{aligned}$$

Let  $m+n=k$ ,

$$x(t)y(t) = \sum_{k=-\infty}^{\infty} \sum_{n=-\infty}^{\infty} a_n b_{k-n} e^{jk\omega_0 t}$$

$$z(t) = x(t)y(t) = \sum_{k=-\infty}^{\infty} c_k e^{jk\omega_0 t}$$

$$\therefore c_k = \sum_{n=-\infty}^{\infty} a_n b_{k-n}$$

$c_k$  could be seen as a convolution of  $a_k$  and  $b_k$ .

b) The period of the signal is 3.  $\omega_0 = \frac{2\pi}{T} = \frac{2\pi}{3}$

Let  $x_1(t) = p(t)q_1(t)$ ,

where  $p(t) = \cos 20\pi t$ ,  $q_1(t) = \begin{cases} 1, & |t| < 1 \\ 0, & 1 < |t| < 1.5 \end{cases}$  and  $q_1(t+3) = q_1(t)$

Since  $p(t) = \cos 20\pi t$

$$= \cos 30\omega_0 t = \frac{1}{2} e^{j30\omega_0 t} + \frac{1}{2} e^{-j30\omega_0 t}$$

$$\therefore a_{30} = \frac{1}{2}, \quad a_{-30} = -\frac{1}{2}$$

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From Table 4.2

$$x(t) = \begin{cases} 1, & |t| < T_1 \\ 0, & T_1 < |t| \leq \frac{T}{2} \end{cases} \quad \text{and } x(t+T) = x(t) \quad \xleftrightarrow{\text{F.S.}} \quad \frac{\sin k\omega_0 T_1}{k\pi}$$

The Fourier Series coefficients of  $q_1(t)$  are,

$$b_k = \frac{\sin k \frac{2\pi}{3}}{k\pi}$$

$$\therefore C_k = a_{30} b_{k-30} + a_{-30} b_{k+30}$$

$$= \frac{1}{2} \cdot \frac{\sin(k-30) \frac{2\pi}{3}}{k\pi} + \frac{1}{2} \frac{\sin(k+30) \frac{2\pi}{3}}{k\pi}$$

ii) The period of the signal is 3.  $\omega_0 = \frac{2\pi}{T} = \frac{2\pi}{3}$ .

Let  $x_2(t) = p(t)q_2(t)$ .

where  $p(t) = \cos 20\pi t$ ,  $q_2(t) = q_2'(t) + q_2''(t)$

where  $q_2'(t) = \begin{cases} 1, & 0 < t < 2 \\ 0, & 2 < t < 3 \end{cases}$  and  $q_2'(t+3) = q_2'(t)$

$q_2''(t) = \begin{cases} 1, & 0 < t < 1 \\ 0, & 1 < t < 3 \end{cases}$  and  $q_2''(t+3) = q_2''(t)$ .

From Table 4.2

$$x(t) = \begin{cases} 1, & |t| < T_1 \\ 0, & T_1 < |t| \leq \frac{T}{2} \end{cases} \quad \text{and } x(t+T) = x(t) \quad \xleftrightarrow{\text{F.S.}} \quad \frac{\sin k\omega_0 T_1}{k\pi}$$

Table 3.1

$$x(t-t_0) \quad \xleftrightarrow{\text{F.S.}} \quad a_k e^{-jk\omega_0 t_0} = a_k e^{-jk \frac{2\pi}{T} t_0}$$

## ECE 301 Signals and Systems Homework # 5 Solution

$$\begin{aligned}
 b_k &= b'_k + b''_k \\
 &= e^{-jk\frac{2\pi}{3} \cdot \frac{1}{2}} \cdot \frac{\sin k\frac{2\pi}{3} \cdot \frac{1}{2}}{k\pi} + e^{-j\frac{2\pi}{3} \cdot 1} \cdot \frac{\sin k\frac{2\pi}{3} \cdot 1}{k\pi} \\
 &= e^{-jk\frac{\pi}{3}} \cdot \frac{\sin k\frac{\pi}{3}}{k\pi} + e^{-j\frac{2\pi}{3}} \cdot \frac{\sin k\frac{2\pi}{3}}{k\pi}
 \end{aligned}$$

$$\begin{aligned}
 C_k &= a_{30} b_{k-30} + a_{-30} b_{k+30} \\
 &= e^{-j(k-30)\frac{\pi}{3}} \cdot \frac{\sin(k-30)\frac{\pi}{3}}{2k\pi} + e^{-j(k-30)\frac{2\pi}{3}} \cdot \frac{\sin(k-30)\frac{2\pi}{3}}{k\pi} \\
 &\quad + e^{-j(k+30)\frac{\pi}{3}} \cdot \frac{\sin(k+30)\frac{\pi}{3}}{k\pi} + e^{-j(k+30)\frac{2\pi}{3}} \cdot \frac{\sin(k+30)\frac{2\pi}{3}}{k\pi}
 \end{aligned}$$

iii). The period of the signal is 4.  $\omega_0 = \frac{2\pi}{T} = \frac{\pi}{2}$ .

Let  $x_3(t) = p(t)q_3(t)$

where  $p(t) = \cos 20\pi t$ ,  $q_3(t) = \begin{cases} e^{-|t|} & , |t| < 1 \\ 0 & , 1 < |t| < 2 \end{cases}$  and  $q_3(t+4) = q_3(t)$

The Fourier Series coefficients of  $q_3(t)$  are

$$\begin{aligned}
 b_k &= \frac{1}{4} \left[ \int_{-1}^0 e^t \cdot e^{-jk\omega_0 t} dt + \int_0^1 e^{-t} \cdot e^{-jk\omega_0 t} dt \right] \\
 &= \frac{1}{4} \left\{ \frac{1}{1-jk\omega_0} e^{(1-jk\omega_0)t} \Big|_{-1}^0 - \frac{1}{1+jk\omega_0} e^{(-1-jk\omega_0)t} \Big|_0^1 \right\} \\
 &= \frac{1}{4} \left\{ \frac{1 - e^{jk\omega_0 - 1}}{1 - jk\omega_0} - \frac{e^{-jk\omega_0 - 1} - 1}{1 + jk\omega_0} \right\}
 \end{aligned}$$

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$$= \frac{1}{4} \cdot \frac{(1 - e^{jk\omega_0 - 1})(1 + jk\omega_0) - (1 - jk\omega_0)(e^{-1 - jk\omega_0} - 1)}{1 + k^2\omega_0^2}$$
$$= \frac{e - \cos(k\omega_0) + k\omega_0 \sin(k\omega_0)}{2e(1 + k^2\omega_0^2)}$$

$$C_k = a_{30} b_{k-30} + a_{-30} b_{k+30}$$

$$= \frac{e - \cos[(k-30)\frac{\pi}{2}] + (k-30)\frac{\pi}{2} \sin[(k-30)\frac{\pi}{2}]}{4e[1 + (k-30)^2 \frac{\pi^2}{4}]}$$
$$+ \frac{e - \cos[(k+30)\frac{\pi}{2}] + (k+30)\frac{\pi}{2} \sin[(k+30)\frac{\pi}{2}]}{4e[1 + (k+30)^2 \frac{\pi^2}{4}]}$$

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C) Let  $z(t) = x(t) \cdot x^*(t)$

$$= |x(t)|^2 = \sum_{k=-\infty}^{\infty} C_k e^{jk\omega_0 t}$$

From Table 3.1

$x^*(t)$	F.S. $\longleftrightarrow$	$b_k = a_{-k}^*$
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$$\begin{aligned} \frac{1}{T_0} \int_0^{T_0} |x(t)|^2 dt &= \frac{1}{T_0} \int_0^{T_0} z(t) dt \\ &= \frac{1}{T_0} \int_0^{T_0} \sum_{k=-\infty}^{\infty} C_k e^{jk\omega_0 t} dt \\ &= \frac{1}{T_0} \left\{ C_0 T_0 + \left[ \sum_{k \neq 0} C_k \cdot \frac{1}{jk\omega_0} \cdot e^{jk\omega_0 t} \right] \Big|_0^{T_0} \right\} \\ &= C_0 \end{aligned}$$

$$\begin{aligned} C_0 &= \sum_{n=-\infty}^{\infty} a_n b_{k-n} \\ &= \sum_{n=-\infty}^{\infty} a_n a_{-n}^* \\ &= \sum_{n=-\infty}^{\infty} |a_n|^2 \end{aligned}$$

$$\therefore \frac{1}{T_0} \int_0^{T_0} |x(t)|^2 dt = \sum_{n=-\infty}^{\infty} |a_n|^2 \quad \text{Parseval's relation}$$



# ECE 301 Signals and Systems Homework # 5 Solution

4.21

a)  $[e^{-at} \cos \omega_0 t] u(t)$  ,  $a > 0$ .

$$\begin{aligned}
X(j\omega) &= \int_{-\infty}^{\infty} [e^{-at} \cos \omega_0 t] u(t) e^{-j\omega t} dt \\
&= \int_0^{\infty} [e^{-at} \cdot \frac{e^{j\omega_0 t} + e^{-j\omega_0 t}}{2}] \cdot e^{-j\omega t} dt \\
&= \frac{1}{2} \int_0^{\infty} e^{[-a-j(\omega-\omega_0)]t} dt + \frac{1}{2} \int_0^{\infty} e^{[-a-j(\omega+\omega_0)]t} dt \\
&= \frac{1}{2} \left. \frac{e^{[-a-j(\omega-\omega_0)]t}}{a+j(\omega-\omega_0)} \right|_0^{\infty} + \frac{1}{2} \cdot \left. \frac{e^{[-a-j(\omega+\omega_0)]t}}{a+j(\omega+\omega_0)} \right|_0^{\infty} \\
&= \frac{1}{2} \cdot \frac{1}{a+j(\omega-\omega_0)} + \frac{1}{2} \cdot \frac{1}{a+j(\omega+\omega_0)}
\end{aligned}$$

Alternatively, you can use Table 4.2

$$e^{-\alpha t} u(t) \xleftrightarrow{\text{F.T.}} \frac{1}{\alpha + j\omega} \quad , \quad \alpha > 0.$$

$$\cos \omega_0 t \xleftrightarrow{\text{F.T.}} \pi [\delta(\omega - \omega_0) + \delta(\omega + \omega_0)]$$

Thus,  $[e^{-at} \cos \omega_0 t] u(t)$

$= e^{-at} u(t) \cdot \cos \omega_0 t$

$\Updownarrow$  F.T.

$\frac{1}{2\pi} \cdot \frac{1}{a+j\omega} * \{ \pi [\delta(\omega - \omega_0) + \delta(\omega + \omega_0)] \}$

$= \frac{1}{2} \cdot \frac{1}{a+j(\omega-\omega_0)} + \frac{1}{2} \cdot \frac{1}{a+j(\omega+\omega_0)}$

## ECE 301 Signals and Systems Homework # 5 Solution

d)  $\sum_{k=0}^{\infty} \alpha^k \delta(t-kT)$  ,  $|\alpha| < 1$ .

From Table 4.1  $\delta(t-t_0) \iff e^{-j\omega t_0}$

$$\sum_{k=0}^{\infty} \alpha^k \delta(t-kT) \xrightarrow{\text{F.T.}} \sum_{k=0}^{\infty} \alpha^k e^{-j\omega kT} = \sum_{k=0}^{\infty} (\alpha e^{-j\omega T})^k = \frac{1}{1 - \alpha e^{-j\omega T}}$$

f) From Table 4.2.

$\frac{\sin \pi t}{\pi t} \xrightarrow{\text{F.T.}} X(j\omega) = \begin{cases} 1, & |\omega| < \pi \\ 0, & |\omega| > \pi \end{cases}$

By the shifting property.

$$\frac{\sin 2\pi(t-1)}{\pi(t-1)} \xrightarrow{\text{F.T.}} X'(j\omega) = \begin{cases} e^{-j\omega} & , |\omega| < 2\pi \\ 0 & , |\omega| > 2\pi \end{cases}$$

$$\left[ \frac{\sin \pi t}{\pi t} \right] \left[ \frac{\sin 2\pi(t-1)}{\pi(t-1)} \right] \xrightarrow{\text{F.T.}} \frac{1}{2\pi} \cdot X(j\omega) * X'(j\omega) = Y(j\omega)$$

i)  $\omega + \pi < -2\pi$  ,  $\omega < -3\pi$

$$Y(j\omega) = 0$$

ii)  $-2\pi < \omega + \pi < 0$  ,  $-3\pi < \omega < -\pi$

$$Y(j\omega) = \frac{1}{2\pi} \int_{-2\pi}^{\omega+\pi} e^{-j\tau} d\omega$$

$$= -\frac{j}{2\pi} (1 + e^{-j\omega})$$

## ECE 301 Signals and Systems Homework # 5 Solution

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iii)  $0 < \omega + \pi < 2\pi$  ,  $-\pi < \omega < \pi$  .

$$Y(j\omega) = \frac{1}{2\pi} \int_{\omega-\pi}^{\omega+\pi} e^{-j\tau} d\tau$$

$$= \frac{1}{2\pi} \cdot (-2 \sin \omega)$$

$$= -\frac{\sin \omega}{\pi}$$

iv)  $2\pi < \omega + \pi < 4\pi$  ,  $\pi < \omega < 3\pi$  .

$$Y(j\omega) = \frac{1}{2\pi} \int_{\omega-\pi}^{2\pi} e^{-j\tau} d\tau$$

$$= \frac{j}{2\pi} (1 + e^{-j\omega})$$

v)  $\omega - \pi > 2\pi$  ,  $\omega > 3\pi$  .

$$Y(j\omega) = 0$$

# ECE 301 Signals and Systems Homework # 5 Solution

i)  $x(t) = \begin{cases} 1-t^2, & 0 < t < 1 \\ 0 & \text{otherwise,} \end{cases}$

$$X(j\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$$
$$= \int_0^1 (1-t^2) e^{-j\omega t} dt \quad (*)$$

$$= \boxed{\frac{1}{j\omega} - \frac{2e^{-j\omega}}{\omega^2} - \frac{2(e^{-j\omega} - 1)}{j\omega^3}}$$

⊗ You will need to do integration by parts twice.

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b)  $X(j\omega) = \cos(4\omega + \frac{\pi}{3})$

$$= \frac{e^{j(4\omega + \frac{\pi}{3})} + e^{-j(4\omega + \frac{\pi}{3})}}{2}$$
$$= \frac{e^{j\frac{\pi}{3}} e^{j4\omega} + e^{-j\frac{\pi}{3}} e^{-j4\omega}}{2}$$

From Table 4.2.

$$\delta(t-t_0) \xleftrightarrow{\text{F.T.}} e^{-j\omega t_0}$$

$$x(t) = \frac{e^{j\frac{\pi}{3}}}{2} \delta(t+4) + \frac{e^{-j\frac{\pi}{3}}}{2} \delta(t-4).$$

### ECE 301 Signals and Systems Homework # 5 Solution

c) 
$$X(j\omega) = \begin{cases} |\omega| e^{-3j\omega} & , |\omega| < 1 \\ 0 & , |\omega| > 1 \end{cases}$$

$$x(t) = \frac{1}{2\pi} \cdot \int_{-1}^0 (-\omega) e^{-3j\omega} e^{j\omega t} d\omega + \frac{1}{2\pi} \int_0^1 \omega e^{-3j\omega} e^{j\omega t} d\omega$$
$$= \frac{1}{\pi} \left[ \frac{\sin(t-3)}{t-3} + \frac{\cos(t-3)-1}{(t-3)^2} \right]$$

d) 
$$X(j\omega) = 2[\delta(\omega-1) - \delta(\omega+1)] + 3[\delta(\omega-2\pi) + \delta(\omega+2\pi)]$$

From Table 4.2.

$\cos \omega_0 t \xLeftrightarrow{\text{F.T.}} \pi [\delta(\omega - \omega_0) + \delta(\omega + \omega_0)]$
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$\sin \omega_0 t \xLeftrightarrow{\text{F.T.}} \frac{\pi}{j} [\delta(\omega - \omega_0) - \delta(\omega + \omega_0)]$
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we obtain.

$x(t) = \frac{2j}{\pi} \sin t + \frac{3}{\pi} \cos(2\pi t)$
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## ECE 301 Signals and Systems Homework # 5 Solution

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$$X_0(t) = \begin{cases} e^{-t} & , 0 \leq t \leq 1 \\ 0 & , 0. \end{cases}$$

$$\begin{aligned} X_0(j\omega) &= \int_{-\infty}^{\infty} X_0(t) e^{-j\omega t} dt \\ &= \int_0^1 e^{-t} e^{-j\omega t} dt \\ &= \frac{1}{-1-j\omega} e^{(-1-j\omega)t} \Big|_0^1 \\ &= \frac{1 - e^{-1-j\omega}}{1+j\omega}. \end{aligned}$$

b)  $X_2(t) = X_0(t) - X_0(-t)$

From Table 4.1

$$X(-t) \iff X(-j\omega)$$

we obtain

$$X_2(j\omega) = \frac{1 - e^{-1-j\omega}}{1+j\omega} - \frac{1 - e^{-1+j\omega}}{1-j\omega}$$

c)  $X_3(t) = X_0(t+1) + X_0(t)$

From Table 4.1

$$X(t-t_0) \iff e^{-j\omega t_0} X(j\omega)$$

$$X_3(j\omega) = \frac{1 - e^{-1-j\omega}}{1+j\omega} + e^{j\omega} \cdot \frac{1 - e^{-1-j\omega}}{1+j\omega}$$

## ECE 301 Signals and Systems Homework # 5 Solution

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4.24

a) (1)  $\text{Re}\{X(j\omega)\} = 0$   
 $\Rightarrow X(j\omega) = -X^*(j\omega),$   
 $\Rightarrow x(t) = -x^*(-t). \quad \text{—————} \quad \textcircled{1}$

For a real signal,  $\textcircled{1}$  becomes  $x(t) = -x(-t)$ , which implies that  $x(t)$  is odd.

Obviously (a) and (d) satisfies the condition.

(2)  $\text{Im}\{X(j\omega)\} = 0$   
 $\Rightarrow X(j\omega) = X^*(j\omega)$   
 $\Rightarrow x(t) = x^*(-t) \quad \text{—————} \quad \textcircled{2}$

For a real signal,  $\textcircled{2}$  becomes  $x(t) = x(-t)$ , which implies that  $x(t)$  is even.

Obviously, (e) and (f) satisfies the condition.

(3) From Table 4.1 we know that  $e^{j\alpha\omega} X(j\omega)$  is the Fourier Transform of  $x(t+\alpha)$ .

If  $e^{j\alpha\omega} X(j\omega)$  is real, from (2) we know that  $x(t+\alpha)$  must be even. So the problem effectively asks whether there exists some real  $\alpha$  that makes  $x(t+\alpha)$  an even signal.

Obviously, (a) and (b) satisfy the condition.

## ECE 301 Signals and Systems Homework # 5 Solution

(4)  $x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega$       (a)(d)(f) satisfy.

Let  $t=0$ ,  $x(0) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) d\omega$ .

$$\boxed{\int_{-\infty}^{\infty} X(j\omega) d\omega = 0 \iff x(0) = 0}$$

(5) From Table 4.1

$$\boxed{\frac{d}{dt} x(t) \iff j\omega X(j\omega)}$$

$$\therefore \frac{d}{dt} x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} j\omega X(j\omega) e^{j\omega t} d\omega$$

Let  $t=0$ ,

(b)(c)(e)(f) satisfy.

$$\frac{d}{dt} x(0) = \frac{j}{2\pi} \int_{-\infty}^{\infty} \omega X(j\omega) d\omega$$

$$\boxed{\int_{-\infty}^{\infty} \omega X(j\omega) d\omega = 0 \iff \frac{d}{dt} x(0) = 0}$$

(6) If  $X(j\omega)$  is periodic, then it can be written in a Fourier Series representation,

$$X(j\omega) = \sum_{k=-\infty}^{\infty} A_k e^{jkt\omega_0}, \quad \omega_0 = \frac{2\pi}{W}, \quad W \text{ is the period of } X(j\omega)$$

From Table 4.2,

$$\boxed{\delta(t-t_0) \xleftrightarrow{\text{F.T.}} e^{j\omega t_0}}$$

$$X(j\omega) = \sum_{k=-\infty}^{\infty} A_k e^{jkt\omega_0} \xleftrightarrow{\text{F.T.}} x(t) = \sum_{k=-\infty}^{\infty} A_k \delta(t - kt_0)$$

It is seen that  $x(t)$  is an impulse train with a period of  $\frac{2\pi}{W}$ .

~~Obviously none of the signals satisfies this.~~ **Signal (b) is of the required form.**



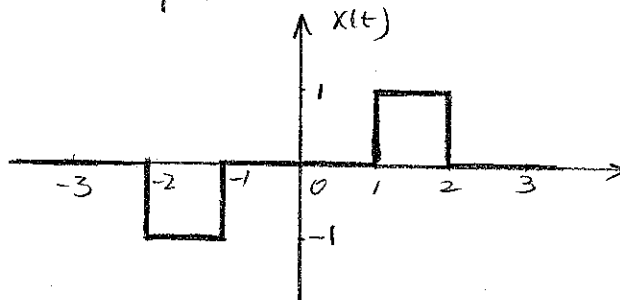
## ECE 301 Signals and Systems Homework # 5 Solution

b) A real signal  $x(t)$  that satisfies (1), (4) and (5) is

such that

$$\begin{cases} x(t) \text{ is odd.} \\ x(0) = 0 \\ \left. \frac{d}{dt} x(t) \right|_{t=0} = 0 \end{cases}$$

One example could be



### ECE 301 Signals and Systems Homework # 5 Solution

4.26 a)

$$x(t) = te^{-2t}u(t) \quad \overset{\text{F.T.}}{\iff} \quad X(j\omega) = \frac{1}{(2+j\omega)^2}$$

$$h(t) = te^{-4t}u(t) \quad \overset{\text{F.T.}}{\iff} \quad H(j\omega) = \frac{1}{(4+j\omega)^2}$$

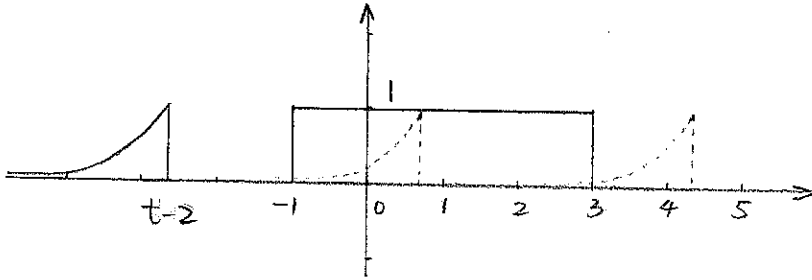
$$\begin{aligned} Y(j\omega) &= \frac{1}{(2+j\omega)^2} \cdot \frac{1}{(4+j\omega)^2} \\ &= \frac{1}{4(2+j\omega)} + \frac{1}{4(2+j\omega)^2} - \frac{1}{4(4+j\omega)} + \frac{1}{4(4+j\omega)^2} \end{aligned}$$

$$y(t) = \frac{1}{4}e^{-2t}u(t) + \frac{1}{4}te^{-2t}u(t) - \frac{1}{4}e^{-4t}u(t) + \frac{1}{4}te^{-4t}u(t)$$

## ECE 301 Signals and Systems Homework # 5 Solution

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### b) Time Domain Convolution



i)  $t-2 < -1$  ,  $t < 1$  .

$$y(t) = 0$$

ii)  $-1 < t-2 < 3$  ,  $1 < t < 5$

$$y(t) = \int_{-1}^{t-2} e^{(\tau-t+2)} d\tau$$

$$= 1 - e^{1-t}$$

iii)  $t-2 > 3$  ,  $t > 5$

$$y(t) = \int_{-1}^3 e^{(\tau-t+2)} d\tau$$

$$= e^{(5-t)} - e^{(1-t)}$$

### Frequency Domain Multiplication

$$x(t) \xleftrightarrow{\text{F.T.}} X(j\omega) = \frac{e^{-j2\omega}}{1+j\omega}$$

$$h(t) \xleftrightarrow{\text{F.T.}} H(j\omega) = e^{-j\omega} \cdot \frac{2 \sin 2\omega}{\omega}$$

$$Y(j\omega) = H(j\omega) X(j\omega)$$

$$= \frac{e^{-j\omega}}{j\omega} - \frac{e^{-j5\omega}}{j\omega} - \frac{e^{-j\omega}}{1+j\omega} + \frac{e^{-j5\omega}}{1+j\omega}$$

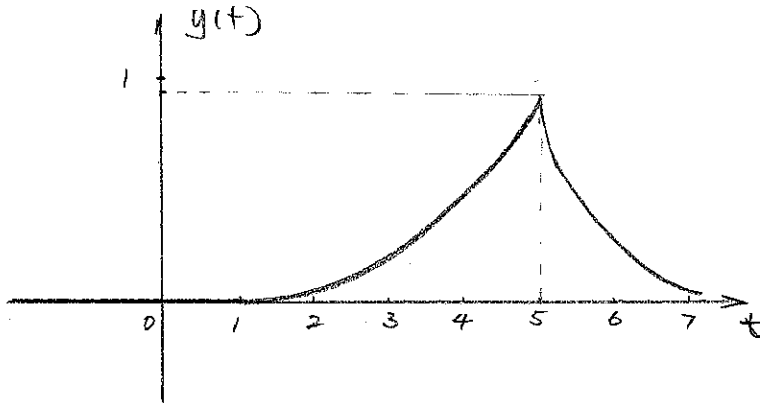
## ECE 301 Signals and Systems Homework # 5 Solution

$$= (e^{-j\omega} - e^{-j5\omega}) \left[ \frac{1}{j\omega} + \pi \delta(\omega) \right] - \frac{1}{2} (e^{-j\omega} - e^{-j5\omega}) 2\pi \delta(\omega) - (e^{-j\omega} - e^{-j5\omega}) \cdot \frac{1}{1+j\omega}$$

The second term  $\frac{1}{2} (e^{-j\omega} - e^{-j5\omega}) 2\pi \delta(\omega)$  is effectively 0.  $e^{-j\omega}$  and  $e^{-j5\omega}$  each correspond to a time shift.

$$Y(j\omega) \xrightarrow{\text{F.T.}} y'(t) = u(t-1) - u(t-5) - [e^{-(t-1)} u(t-1) - e^{-(t-5)} u(t-5)]$$

It can be shown that  $y'(t) = y(t)$ .



## ECE 301 Signals and Systems Homework # 5 Solution

4.28 a) From Table 4.2

$$e^{j\omega_0 t} \xleftrightarrow{\text{F.T.}} 2\pi\delta(\omega - \omega_0)$$

$$P(j\omega) = \sum_{n=-\infty}^{\infty} a_n \cdot 2\pi\delta(\omega - n\omega_0)$$

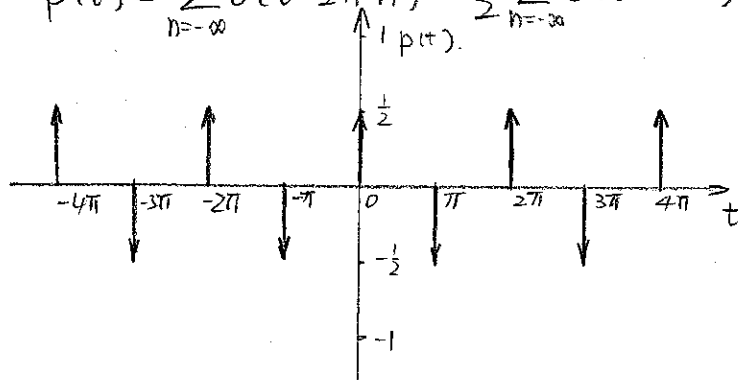
$$\begin{aligned} \therefore y(t) = x(t) \cdot p(t) &\xleftrightarrow{\text{F.T.}} \frac{1}{2\pi} X(j\omega) \cdot P(j\omega) \\ &= \frac{1}{2\pi} X(j\omega) \cdot \sum_{n=-\infty}^{\infty} a_n \cdot 2\pi\delta(\omega - n\omega_0) \\ &= \sum_{n=-\infty}^{\infty} a_n X[j(\omega - n\omega_0)] \end{aligned}$$

b) iii)  $p(t) = \cos(2t) = \frac{e^{j2t} + e^{-j2t}}{2}$

$\therefore a_1 = a_{-1} = \frac{1}{2}, \omega_0 = 2.$

$\therefore Y(j\omega) = \frac{1}{2} X[j(\omega - 2)] + \frac{1}{2} X[j(\omega + 2)].$

ix)  $p(t) = \sum_{n=-\infty}^{\infty} \delta(t - 2\pi \cdot n) - \frac{1}{2} \sum_{n=-\infty}^{\infty} \delta(t - n\pi)$



$$a_k = \int_{-\epsilon}^{2\pi - \epsilon} \frac{1}{2} [\delta(t) - \delta(t - \pi)] e^{-jkt} dt.$$

$$= \frac{1}{2} (1 - e^{-jk\pi}) = \frac{1 - (-1)^k}{2}$$

$$Y(j\omega) = \sum_{n=-\infty}^{\infty} a_n X[j(\omega - n)] = \frac{1}{2} \sum_{n=-\infty}^{\infty} X[j(\omega - 2n - 1)]$$

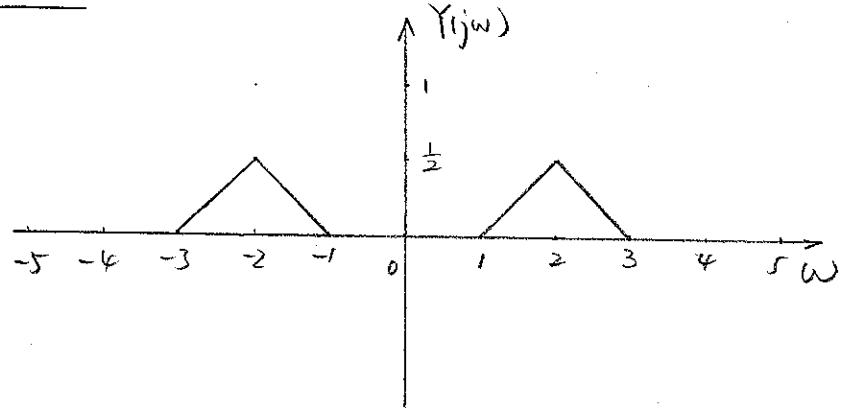
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# ECE 301 Signals and Systems

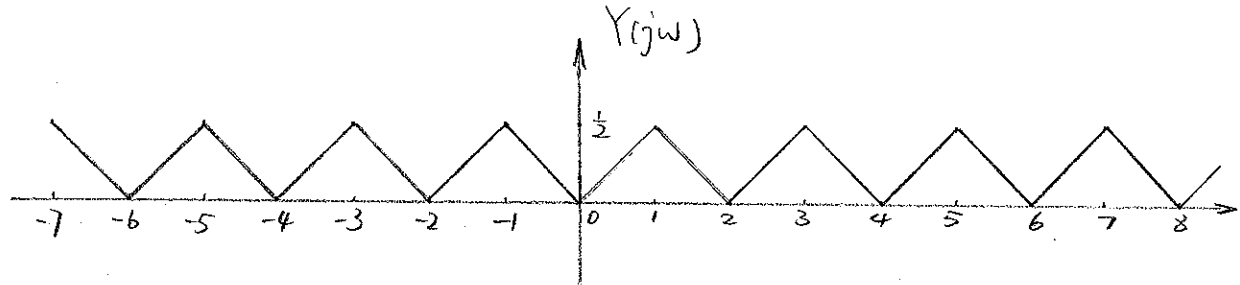
## Homework # 5 Solution

Spectrum

iii)



ix)



# ECE 301 Signals and Systems

## Homework # 5 Solution

4.33

b)

$$\frac{d^2 y(t)}{dt^2} + 6 \frac{dy(t)}{dt} + 8y(t) = 2x(t)$$

↕ F.T.

$$(j\omega)^2 Y(j\omega) + 6j\omega Y(j\omega) + 8Y(j\omega) = 2X(j\omega)$$

$$Y(j\omega) = \frac{2}{(j\omega)^2 + 6j\omega + 8} \cdot X(j\omega)$$

$$X(j\omega) = \frac{1}{(2+j\omega)^2}$$

$$\therefore Y(j\omega) = \frac{2}{(2+j\omega)(4+j\omega)} \cdot \frac{1}{(2+j\omega)^2}$$

$$= \frac{2}{(4+j\omega)(2+j\omega)^3}$$

$$= -\frac{1}{4(4+j\omega)} + \frac{1}{4(2+j\omega)} - \frac{1}{2(2+j\omega)^2} + \frac{1}{(2+j\omega)^3}$$

↕ F.T.

$$y(t) = -\frac{1}{4} e^{-4t} u(t) + \frac{1}{4} e^{-2t} u(t) - \frac{1}{2} t e^{-2t} u(t) + \frac{t^2}{2} e^{-2t} u(t)$$