

## Fourier Series

### LTI systems & complex exponentials

We have shown that we can always express the output of an LTI system as a convolution:

$$y(t) = \int_{-\infty}^{\infty} h(\tau) x(t-\tau) d\tau$$

$$y[n] = \sum_{k=-\infty}^{\infty} h[k] x[n-k]$$

Let's let the input be a complex exponential:  $e^{j\omega t}$  or  $e^{j\omega n}$   
What is the output:

Continuous - Time:

$$\begin{aligned} y(t) &= \int_{-\infty}^{\infty} h(\tau) x(t-\tau) d\tau \\ &= \int_{-\infty}^{\infty} h(\tau) e^{j\omega(t-\tau)} d\tau \\ &= e^{j\omega t} \underbrace{\int_{-\infty}^{\infty} h(\tau) e^{-j\omega\tau} d\tau}_{H(j\omega)} = H(j\omega) e^{j\omega t} \end{aligned}$$

$H(j\omega)$  depends only on the impulse response and frequency ( $\omega$ )

Discrete - Time:

$$\begin{aligned} y[n] &= \sum_{k=-\infty}^{\infty} h[k] x[n-k] \\ &= \sum_{k=-\infty}^{\infty} h[k] e^{j\omega(n-k)} \\ &= \underbrace{\sum_{k=-\infty}^{\infty} h[k] e^{-j\omega k}}_{H(j\omega)} e^{j\omega n} \end{aligned}$$

Big picture Result: If you put in a signal with frequency  $\omega$  to an LTI system, the output is also of frequency  $\omega$ !

# Harmonically Related Complex Exponentials

## Continuous - Time

Let  $\phi_k(t) = e^{jk\omega_0 t}$  be a family of complex exponentials that is fixed for a given  $\omega_0$

Ex:  $\omega_0 = 100$

$\phi_0(t) = 1$

Ex  $\phi_1(t) = e^{j100t}$

$\phi_{-1}(t) = e^{-j100t}$  etc.

Property 1: All of  $\phi_k$ 's have the same period

Let  $T = \frac{2\pi}{\omega_0}$

$$\begin{aligned}\phi_k(t+T) &= e^{jk\omega_0(t+T)} \\ &= (e^{jk\omega_0 t})(e^{jk\omega_0 T}) \\ &= (e^{jk\omega_0 t})(e^{jk\omega_0 \frac{2\pi}{\omega_0}}) = e^{jk\omega_0 t} = \phi_k(t)\end{aligned}$$

Property 2:  $\phi_k$  is orthogonal to  $\phi_l$

$$\begin{aligned}\int_T \phi_k(t) \phi_l^*(t) dt &= \int_{t_0}^{t_0+T} e^{jk\omega_0 t} e^{-jl\omega_0 t} dt \\ &= \int_{t_0}^{t_0+T} e^{j\omega_0 t(k-l)} dt = \frac{1}{j\omega_0(k-l)} e^{j\omega_0 t(k-l)} \Big|_{t_0}^{t_0+T} \\ &= \frac{1}{j\omega_0(k-l)} (e^{j\omega_0(k-l)(t_0+T)} - e^{j\omega_0(k-l)t_0}) \\ &= \frac{1}{j\omega_0(k-l)} e^{j\omega_0(k-l)t_0} (e^{j(k-l)\omega_0 \frac{2\pi}{\omega_0}} - 1) = 0\end{aligned}$$

So, for any  $k \neq l$   $\int_T \phi_k(t) \phi_l^*(t) dt = 0$   
unique

Property 3: There are an infinite number of  $\phi_k$ 's since  $k$  can be any integer and each  $\phi_k$  is orthogonal from all the others.

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Periodic Signals can be expressed as linear combinations of Harmonically Related Complex Exponentials:

$$x(t) = \sum_{k=-\infty}^{\infty} a_k e^{jk \frac{2\pi}{T} t} = \sum_{k=-\infty}^{\infty} a_k \phi_k(t)$$

since each  $\phi_k(t)$  has period  $T$ ,  $x(t)$  is periodic with period  $T$ .

Finding  $a_k$ :

$$\int_T x(t) e^{-j \frac{2\pi}{T} \ell t} dt = \int_T \sum_{k=-\infty}^{\infty} a_k \phi_k(t) \phi_{\ell}^*(t) dt$$

$$= \sum_{k=-\infty}^{\infty} a_k \int_T \phi_k(t) \phi_{\ell}^*(t) dt$$

Remember, if  $k \neq \ell$   $\int_T \phi_k(t) \phi_{\ell}^*(t) dt = 0$

If  $k = \ell$   $\int_T \phi_k(t) \phi_{\ell}^*(t) dt = \int_T e^{-jk \frac{2\pi}{T} t} e^{jk \frac{2\pi}{T} t} dt = T$

$$= \sum_{k=-\infty}^{\infty} a_k T \delta(k - \ell)$$

$$= T a_{\ell}$$

so  $a_{\ell} = \frac{1}{T} \int_T x(t) e^{-j \frac{2\pi}{T} \ell t} dt$

Fourier Series Equations:

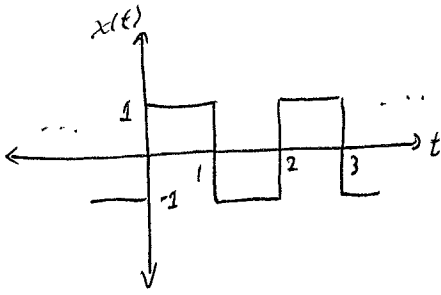
★  $x(t) = \sum_{k=-\infty}^{\infty} a_k e^{jk \frac{2\pi}{T} t}$

$a_k = \frac{1}{T} \int_T x(t) e^{-j \frac{2\pi}{T} k t} dt$  ★

## Intuition

Each harmonic (value of  $\pm k$ ) combines to fill out shape of the signal.

Ex:



period  $T=2$

$$\omega_0 = \frac{2\pi}{T} = \frac{2\pi}{2} = \pi$$

$$a_k = \frac{1}{T} \int_T x(t) e^{-jk \frac{2\pi}{T} t} dt$$

$$= \frac{1}{2} \int_0^2 x(t) e^{-jk\pi t} dt$$

$$= \frac{1}{2} \left[ \int_0^1 (1) e^{-jk\pi t} dt + \int_1^2 (-1) e^{-jk\pi t} dt \right]$$

$$= \frac{1}{2} \left[ \frac{-1}{jk\pi} e^{-jk\pi t} \Big|_0^1 + \frac{1}{jk\pi} e^{-jk\pi t} \Big|_1^2 \right]$$

$$= \frac{1}{2} \frac{1}{jk\pi} \left[ -(e^{-jk\pi} - 1) + (e^{-jk2\pi} - e^{-jk\pi}) \right]$$

$$= \frac{1}{j2k\pi} \left( -(-1)^k + 1 + 1 - (-1)^k \right) = \frac{1 - (-1)^k}{jk\pi} = \begin{cases} 0 & k \text{ even} \\ \frac{2}{jk\pi} & k \text{ odd} \end{cases}$$

$$\begin{aligned} \text{Remember, } x(t) &= \sum_{k=-\infty}^{\infty} a_k e^{jk \frac{2\pi}{T} t} \\ &= \sum_{k=-\infty}^{\infty} \frac{1 - (-1)^k}{jk\pi} e^{jk\pi t} \end{aligned}$$

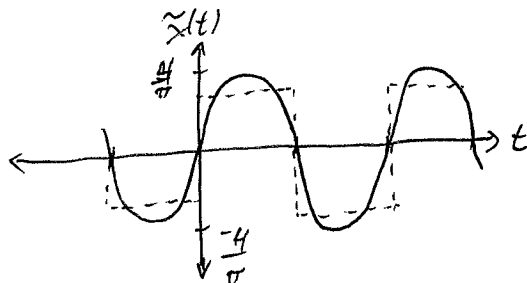
Let's see what happens as we use more and more harmonics to construct  $x(t)$ .

(more harmonics  $\Leftrightarrow$  higher values of  $k$ )

$$\tilde{x}(t) = \sum_{k=-N}^N \frac{1 - (-1)^k}{jk\pi} e^{jk\pi t}$$

$N=1$ :

$$\begin{aligned}\tilde{x}(t) &= \frac{2}{j1\pi} e^{jk\pi t} + \frac{2}{j(-1)\pi} e^{j(-1)\pi t} \\ &= \frac{2}{j\pi} (e^{j\pi t} - e^{-j\pi t}) = \frac{2}{j\pi} 2j \sin(\pi t) = \frac{4}{\pi} \sin(\pi t)\end{aligned}$$



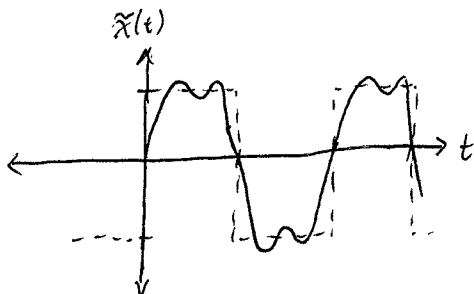
Already, we can see that

$$\tilde{x}(t) \approx x(t)$$

As we add more harmonics,  $\tilde{x}(t)$  will get closer and closer to  $x(t)$

$N=3$ :

$$\begin{aligned}\tilde{x}(t) &= \frac{4}{\pi} \sin(\pi t) + \frac{32}{j3\pi} e^{-j3\pi t} + \frac{2}{j(-3)\pi} e^{j(-3)\pi t} \\ &= \frac{4}{\pi} \sin(\pi t) + \frac{2}{j3\pi} (e^{j3\pi t} - e^{-j3\pi t}) \\ &= \frac{4}{\pi} \sin(\pi t) + \frac{2}{j3\pi} 2j \sin(3\pi t) \\ &= \frac{4}{\pi} \sin(\pi t) + \frac{4}{3\pi} \sin(3\pi t)\end{aligned}$$

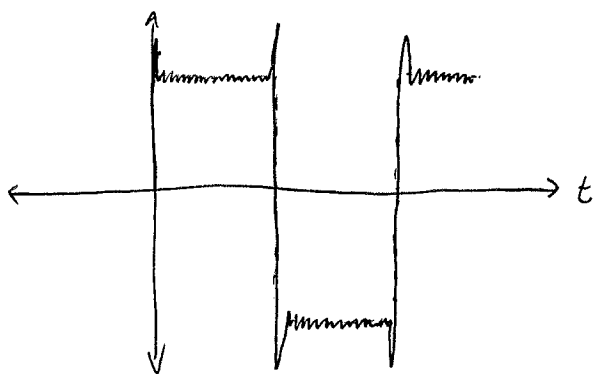


$E_x$

as the number of harmonics goes to  $\infty$ ,  $\tilde{x}(t)$  converges to  $x(t)$  except for locations of discontinuities.

## Gibbs Phenomenon:

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At discontinuities you get overshoot beyond  $x(t)$ .

We're engineers, so we neglect this issue.

## Discrete - Time

### Harmonically Related Complex Exponentials

Let  $\phi_k[n] = e^{jk\omega_0 n}$  be a family of complex exponentials that is fixed for a given  $\omega_0$

Ex:  $\omega_0 = \frac{2\pi}{5}$

$\phi_0[n] = 1$

Ex:  $\phi_1[n] = e^{j\frac{2\pi}{5}n}$

$\phi_{-1}[n] = e^{-j\frac{2\pi}{5}n}$

Property 1: All of  $\phi_k[n]$ 's have same period

Let  $N = 5$   $\omega_0 = \frac{2\pi}{5}$

$$\begin{aligned}\phi_k[n+N] &= \phi_k e^{jk\omega_0[n+N]} = e^{jk\frac{2\pi}{5}n} e^{jk\frac{2\pi}{5}5} \\ &= e^{jk\frac{2\pi}{5}n} (1) = \phi_k[n]\end{aligned}$$

Property 2:  $\phi_k[n]$  is orthogonal to  $\phi_l[n]$  for  $k \neq l$

$$\begin{aligned}\sum_{n=0}^{N-1} \phi_k[n] \phi_l^*[n] &= \sum_{n=0}^{N-1} e^{jk\omega_0 n} e^{-jl\omega_0 n} \\ &= \sum_{n=0}^{N-1} e^{j(k-l)\omega_0 n} = \frac{1 - e^{j(k-l)\omega_0 N}}{1 - e^{j(k-l)\omega_0}}\end{aligned}$$

$$= \frac{1 - e^{j(k-l)\frac{2\pi}{N}N}}{1 - e^{j(k-l)\omega_0}} = 0$$

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Property 3: There are only  $N$  (finite!) unique  $\phi_k[n]$ 's.

Let's compare  $\phi_{k+N}[n]$  &  $\phi_k[n]$

$$\begin{aligned}\phi_{k+N}[n] &= e^{j(k+N)\omega_0 n} \\ &= e^{jk\omega_0 n} e^{jN\frac{2\pi}{N}n} = e^{jk\omega_0 n} = \phi_k[n]\end{aligned}$$

So,  $\phi_{k+N}[n]$  and  $\phi_k[n]$  are NOT different.

Ex:  $N=5$

$$\phi_1[n] = e^{j\frac{2\pi}{5}n}$$

$$\phi_6[n] = e^{j\frac{2\pi}{5}(6)n} = e^{j\frac{12\pi}{5}n}$$

$$\stackrel{\text{Ex}}{=} e^{j(\frac{12\pi}{5} - \frac{10\pi}{5})n} = e^{j\frac{2\pi}{5}n}$$

Periodic signals can be represented as a linear combination of harmonically related complex exponentials.

$$x[n] = \sum_{k=0}^{N-1} a_k e^{j\frac{2\pi}{N}kn} = \sum_{k=0}^{N-1} a_k \phi_k[n]$$

since each  $\phi_k[n]$  is periodic with period  $N$ ,

$x[n]$  is also periodic with period  $N$ .

Finding  $a_k$ :

$$\sum_{n=0}^{N-1} x[n] e^{-j\ell \frac{2\pi}{N} n} = \sum_{n=0}^{N-1} \sum_{k=0}^{N-1} a_k e^{jk \frac{2\pi}{N} n} e^{-j\ell \frac{2\pi}{N} n}$$

$$= \sum_{k=0}^{N-1} a_k \sum_{n=0}^{N-1} e^{jk \frac{2\pi}{N} n} e^{-j\ell \frac{2\pi}{N} n}$$

$$= \sum_{k=0}^{N-1} a_k \sum_{n=0}^{N-1} \phi_k[n] \phi_\ell^*[n]$$

Remember, if  $k \neq \ell$   $\sum_{n=0}^{N-1} \phi_k[n] \phi_\ell^*[n] = 0$

If  $k = \ell$   $\sum_{n=0}^{N-1} \phi_k[n] \phi_\ell^*[n] = \sum_{n=0}^{N-1} e^{j \frac{2\pi}{N} k n} e^{-j \frac{2\pi}{N} k n} = N$

$$\rightarrow = \sum_{k=0}^{N-1} a_k N \delta[k - \ell]$$

$$= N a_\ell$$

$$\text{so, } a_k = \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{-j \frac{2\pi}{N} k n}$$

Fourier Series Equations

$$x[n] = \sum_{k=0}^{N-1} a_k e^{jk \frac{2\pi}{N} n}$$

$$a_k = \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{-j \frac{2\pi}{N} k n}$$

Intuition:

The discrete-time Fourier Series works exactly as the continuous-time: adding more harmonics gets you closer to the actual original signal,  $x[n]$

Also, there is no Gibbs Phenomenon because there are no discontinuities in discrete-time.

This also means that you only need a finite number of terms to make  $x[n]$  exactly.