

key relationship btw DT freq and CT freq:

$$\omega_d = \frac{\omega_a}{F_s}$$

- Sampling on Analog Sinewave

$$\underline{x_a(t)} = e^{j\omega_a t}, \quad \omega_a: \text{analog freq in radians/sec}$$

$$F_s = \frac{1}{T_s} = \text{number of samples/sec (sampling rate in Hz)}$$

$$\begin{aligned} \underline{x[n]} &= x_a(nT_s) = x_a\left(\frac{n}{F_s}\right) \\ &= e^{j\omega_a nT_s} = \underline{e^{j\omega_d n}} \end{aligned}$$

where $\omega_d = \frac{\omega_a}{F_s} = \omega_a T_s$, n : integer

↑ discrete time freq.

- Consider $x_{ae}(t) = e^{j(\omega_a t + l 2\pi F_s) t}$, l : integer

$$\begin{aligned}
 x_e[n] &= x_{ae}(nT_s) = x_{ae}\left(\frac{n}{F_s}\right) \\
 &= \frac{e^{j\omega_a n T_s}}{e^{j\omega_a n T_s}} e^{j l 2\pi F_s n T_s} \\
 &= e^{j\omega_d n} \quad e^{j l 2\pi n} = 1 \\
 &= e^{j\omega_d n}, \quad \omega_d = \frac{\omega_a}{F_s}
 \end{aligned}$$

$e^{j\omega_a t} \xrightarrow{\text{sample}} e^{j\omega_d n}$
 $e^{j(\omega_a + l 2\pi F_s)t} \rightarrow e^{j\omega_d n}$

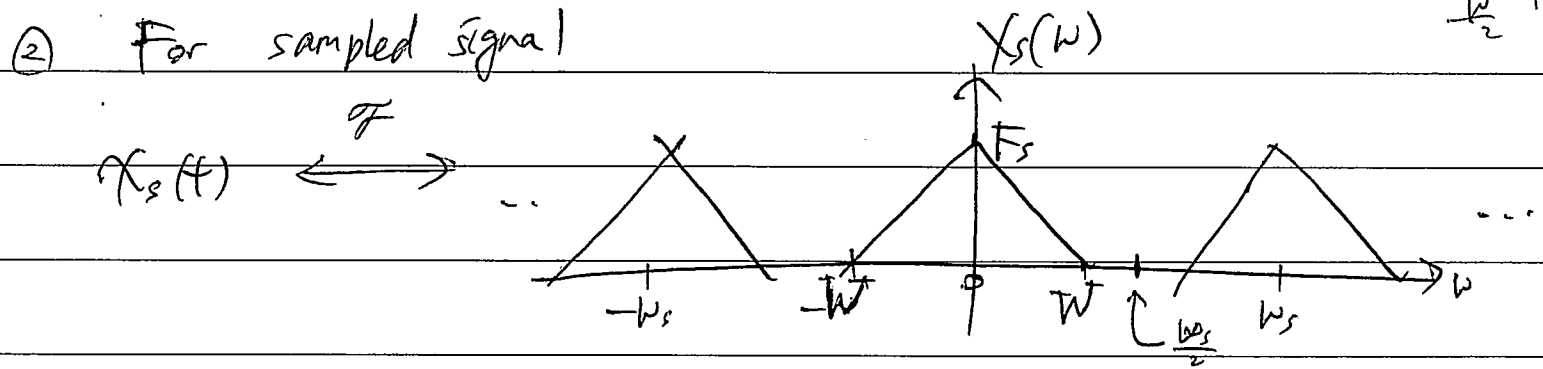
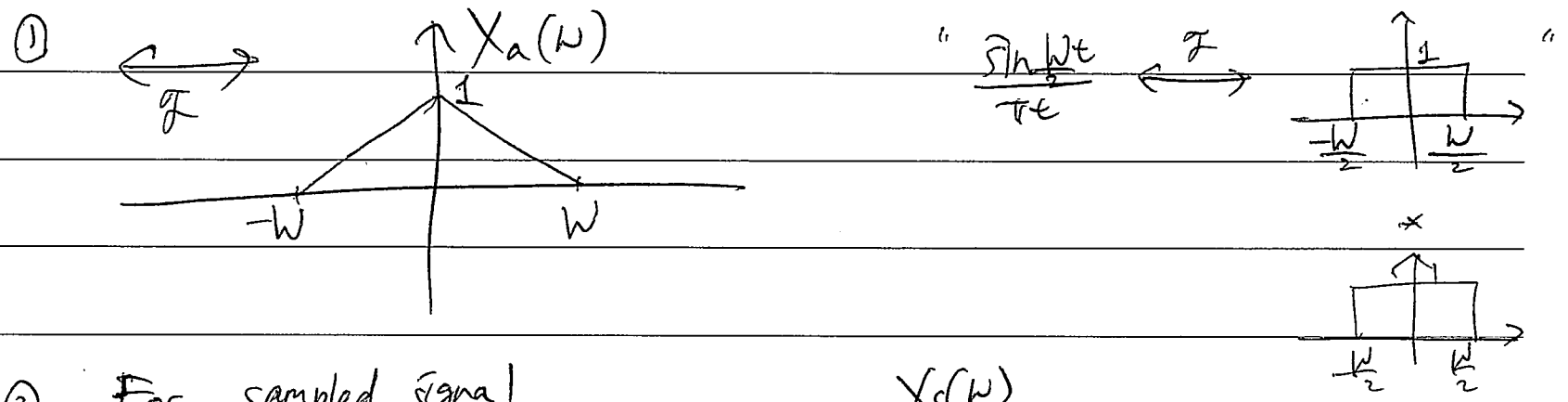
↳ any two analog freq separated by $l\omega_s (= l 2\pi F_s)$ get mapped to same DT sinewave

↳ since $\omega_d = \frac{\omega_a}{F_s}$, any 2 CT sinewaves separated by $l\omega_s$ yields 2 DT sinewaves separated by $l 2\pi$ (same DT sinewave)

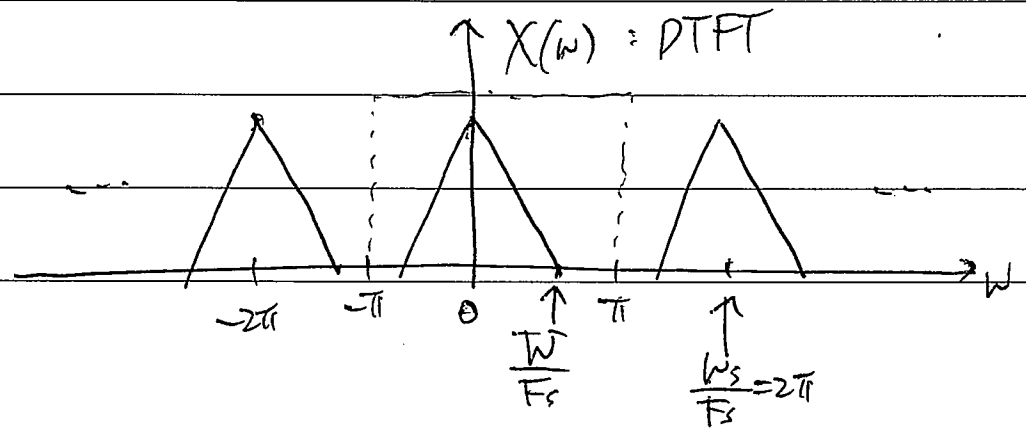
$\frac{l\omega_s}{F_s} = \frac{l 2\pi F_s}{F_s} = l 2\pi$

Example $X_a(t) = \frac{2\pi}{W} \left\{ \frac{\sin(\frac{W}{2}t)}{\pi t} \right\}^2$

sampled at a rate $\omega_s > 2\omega_{max}$, where $\omega_{max} = W$



③ Compress / divide by sampling rate $F_s = \frac{1}{T_s}$

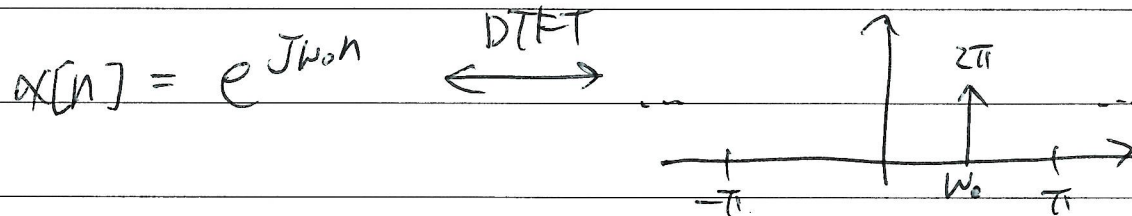
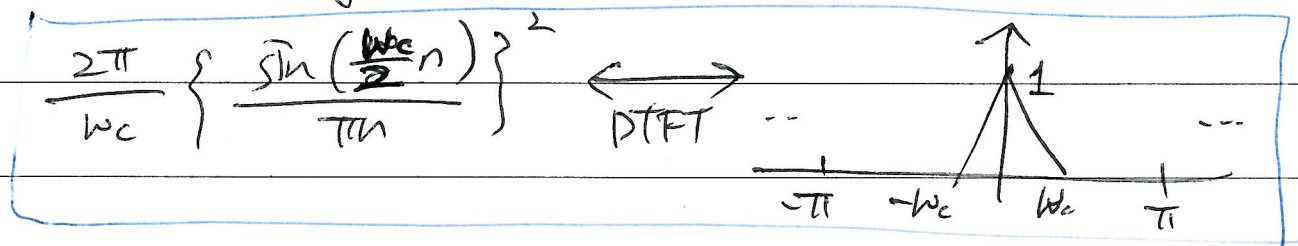


Note: $x[n] = X_a(nT_s) = \frac{2\pi}{W} \left\{ \frac{\sin(\frac{W}{2} nT_s)}{\pi nT_s} \right\}^2$

$$= \frac{2\pi}{W} \left\{ \frac{\sin(\frac{W}{2F_s} n)}{\pi n/F_s} \right\}^2 = \frac{2\pi F_s^2}{W} \left\{ \frac{\sin(\frac{W_c}{2} n)}{\pi n} \right\}^2$$

where: $W_c = \frac{W}{F_s}$

Now divide by $F_s \Rightarrow$ divide by F_s in freq domain and we get our first DTFT pair



DTFT : Properties & Examples.

• DTFT $X(\omega) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n}$

• I-DTFT $x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(\omega) e^{j\omega n} d\omega$

(Chap. 4)

(Chap. 5)

* Two Primary differences btw CTFT and DTFT

- In DT domain, n is a discrete variable

- DTFT is periodic with period 2π

↳ As a result, some of the properties of the DTFT are quite different than the properties of the CTFT, although some are similar.

• Similar properties $x[n] \xleftrightarrow{\text{DTFT}} X(\omega)$

- time-shift $x[n-n_0] \xleftrightarrow{\text{DTFT}} e^{-j\omega n_0} X(\omega)$

- convolution $y[n] = x[n] * h[n] \xleftrightarrow{\text{DTFT}} Y(\omega) = H(\omega) X(\omega)$

- modulation $e^{j\omega_0 n} x[n] \xleftrightarrow{\text{DTFT}} X(\omega - \omega_0)$

but need to shift all periods of $X(\omega)$ to the right by ω_0 (example shortly)

• Different properties.

• Since n is discrete while ω is continuous, there is no duality property for DTFT.

- Since n must be an integer, the time-scaling or freq-scaling property is very different.

(recall) $x(at) \xleftrightarrow{\text{CTFT}} \frac{1}{|a|} X\left(\frac{\omega}{a}\right)$, true for any a .

For DTFT, we have two corresponding properties.

$$\textcircled{1} \quad y[n] = x[Dn] \quad \xleftrightarrow{\text{DTFT}} \quad Y(\omega) = \frac{1}{D} \sum_{k=0}^{D-1} X\left(\frac{\omega - k2\pi}{D}\right)$$

, where $D = \text{integer} > 1$

$$\textcircled{2} \quad y[n] = \begin{cases} x\left[\frac{n}{L}\right], & n = dL \\ 0, & \text{o/w} \end{cases} \quad \xleftrightarrow{\text{DTFT}} \quad Y(\omega) = X(L\omega)$$

, where $L, d = \text{integer}$

↳ These two prop. are beyond the scope of 301,
 so you will not be tested on them.

(time-d)

(freq-d)

Conti - signal

CTFT → aperiodic

discrete signal

DTFT → periodic