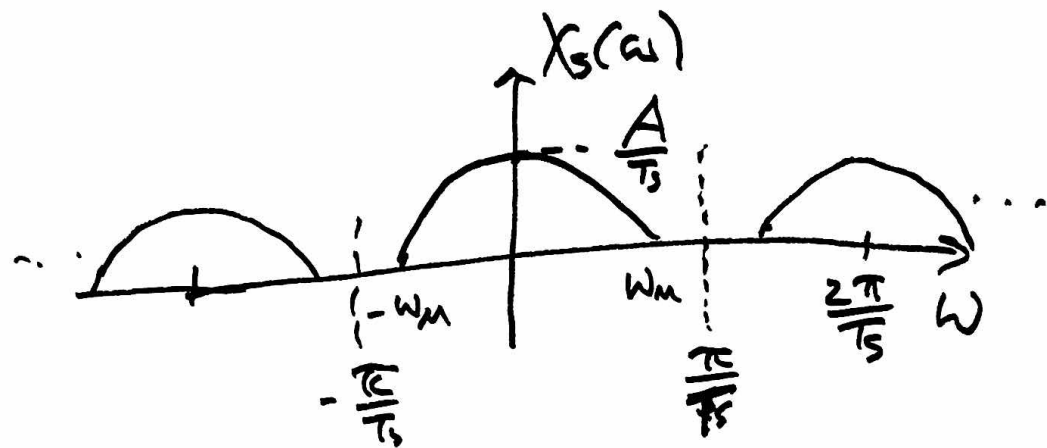
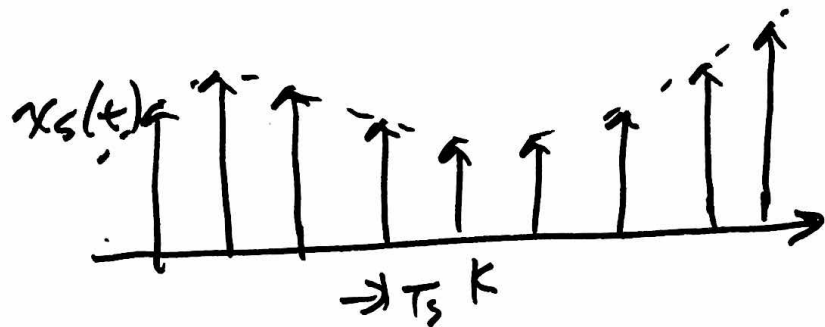
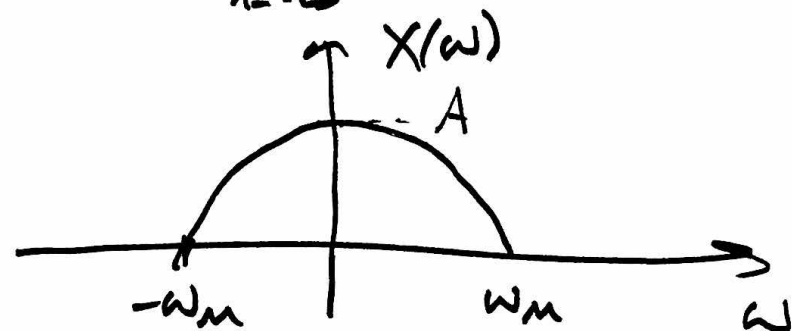
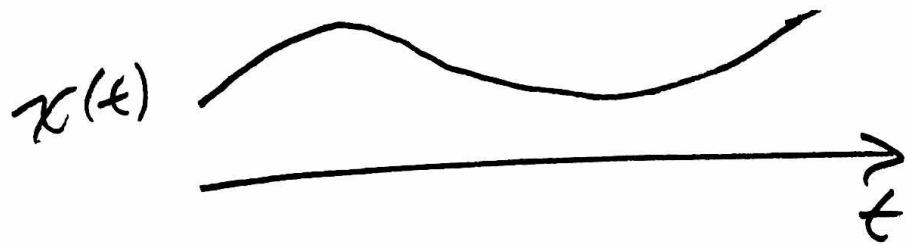


Discrete Time Fourier Transform (DTFT)

→ Going from sampling

• Let $x(t)$ be a bandlimited signal ($X(\omega) = 0$ $\omega > \omega_m$)

$$\text{and } x_s(t) = x(t) \cdot \sum_{n=-\infty}^{\infty} \delta(t - nT_s) = x_s(t) = \sum_{n=-\infty}^{\infty} x(nT_s) \delta(t - nT_s)$$



$$\text{Call } x(nT_s) = x[n]$$

Look at the synthesis for $x[n]$:

$$x[n] = x(nT_s) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) e^{+j\omega n T_s} d\omega = \frac{1}{2\pi} \int_{-\frac{\pi}{T_s}}^{\frac{\pi}{T_s}} X_S(\omega) \cdot T_s e^{+j\omega n T_s} d\omega$$

In order to "untie" the dependence on time, we'll change units:

$$\omega : \frac{\text{rad}}{\text{sec}} \quad \phi : \frac{\text{rad}}{\text{sample}}$$

$$\frac{\text{rad}}{\text{sec}} \Big| \frac{T_s \text{ sec}}{\text{sample}} \rightarrow \omega T_s = \phi$$

$$\omega = \frac{\phi}{T_s}$$

$$d\omega = \frac{1}{T_s} d\phi$$

$$x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X_S\left(\frac{\phi}{T_s}\right) T_s e^{+j\left(\frac{\phi}{T_s}\right) n T_s} \frac{1}{T_s} d\phi$$

$$= \frac{1}{2\pi} \int_{-\pi}^{\pi} X_S\left(\frac{\phi}{T_s}\right) e^{+j\phi n} d\phi \Rightarrow \text{No longer depends on seconds, just samples.}$$

Looking at the analysis equation:

$$X_s(\omega) = \int_{-\infty}^{\infty} x_s(t) e^{-j\omega t} dt = \int_{-\infty}^{\infty} \left(\sum_{n=-\infty}^{\infty} x(nT_s) \delta(t - nT_s) \right) e^{-j\omega t} dt$$

$$= \sum_{n=-\infty}^{\infty} x(nT_s) \left(\int_{-\infty}^{\infty} \delta(t - nT_s) e^{-j\omega t} dt \right)$$

$$= \sum_{n=-\infty}^{\infty} x(nT_s) e^{-j\omega nT_s} \int_{-\infty}^{\infty} \delta(t - nT_s) dt$$

$$= \sum_{n=-\infty}^{\infty} x(nT_s) e^{-j\omega nT_s}$$

$$= \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega nT_s}$$

Normalize frequency: $\omega T_s = \phi$

$$\omega = \frac{\phi}{T_s}$$

$$X_s\left(\frac{\phi}{T_s}\right) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\phi n}$$

From now on, we will use $X(\omega)$ when talking about the DTFT, not $X_s(\frac{\omega}{T_s})$. Context will tell whether ω is $\frac{\text{rad}}{\text{sec}}$ or $\frac{\text{rad}}{\text{sample}}$.

CTFT
 $X(\omega)$

DTFT
 $X(\omega)$

$$X_s\left(\frac{\omega}{T_s}\right) \rightarrow X(\omega)$$

(not the same $X(\omega)$ from page 1)

DTFT equations

$$X(\omega) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n}$$

$$x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(\omega) e^{j\omega n} d\omega$$

$$X(\omega + 2\pi l) = \sum_{n=-\infty}^{\infty} x[n] e^{-j(\omega + 2\pi l)n}$$

$$= \sum_n x[n] e^{-j\omega n} e^{-j2\pi l n}$$

$$= \sum_n x[n] e^{-j\omega n}$$

Linearity & Time shift

$$\rightarrow \text{DTFT}\{ax[n] + by[n]\} = a \text{DTFT}\{x[n]\} + b \text{DTFT}\{y[n]\}$$

$$\rightarrow x[n] \xleftrightarrow{\text{DTFT}} X(\omega)$$
$$x[n-n_0] \xleftrightarrow{\text{DTFT}} e^{-j\omega n_0} X(\omega)$$

Ex

$$y[n] - 0.5y[n-1] = x[n] + 2x[n-1]$$

Find $H(\omega) = \frac{Y(\omega)}{X(\omega)}$

$$y[n] \xleftrightarrow{\text{DTFT}} Y(\omega)$$
$$x[n] \xleftrightarrow{\text{DTFT}} X(\omega)$$

$$Y(\omega) - 0.5Y(\omega)e^{-j\omega} = X(\omega) + 2X(\omega)e^{-j\omega}$$

$$Y(\omega)(1 - 0.5e^{-j\omega}) = X(\omega)(1 + 2e^{-j\omega})$$

$$H(\omega) = \frac{Y(\omega)}{X(\omega)} = \frac{1 + 2e^{-j\omega}}{1 - 0.5e^{-j\omega}}$$

\rightarrow Frequency response of the system

Ex $x[n] = \begin{cases} 1, & 0 \leq n \leq N-1 \\ 0, & \text{else} \end{cases}$

Find $X(\omega)$:

Apply the DTFT analysis eqn:

$$X(\omega) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n}$$

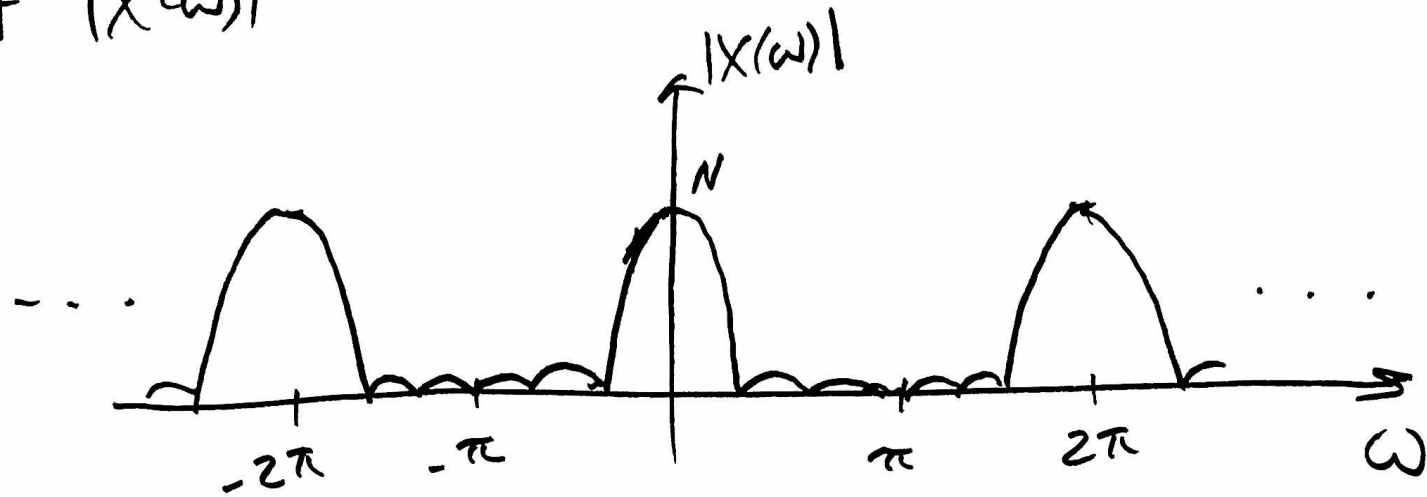
$$= \sum_{n=0}^{N-1} 1 \cdot e^{-j\omega n} = \frac{1 - e^{-j\omega N}}{1 - e^{-j\omega}}$$

$$= \frac{e^{-j\frac{\omega N}{2}}}{e^{-j\frac{\omega}{2}}} \frac{e^{j\frac{\omega N}{2}} - e^{-j\frac{\omega N}{2}}}{e^{j\frac{\omega}{2}} - e^{-j\frac{\omega}{2}}}$$

$$= e^{-j\frac{\omega}{2}(N-1)} \frac{\sin\left(\frac{\omega N}{2}\right)}{\sin\left(\frac{\omega}{2}\right)}$$

Looks like a periodic sinc.
Sometimes called psinc

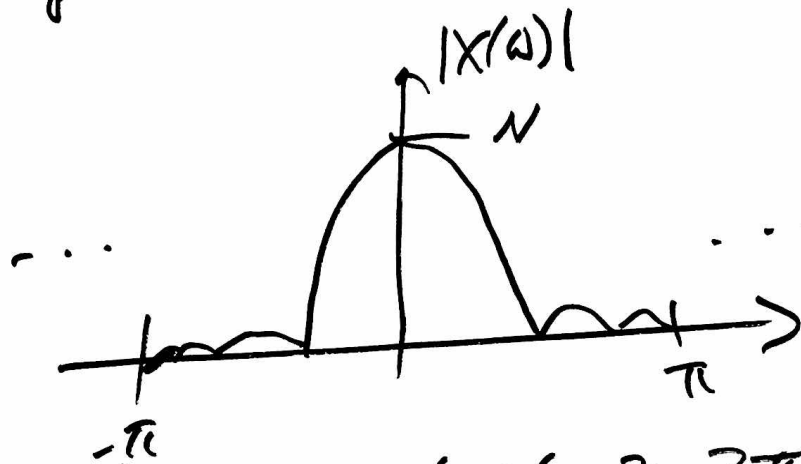
Plot $|X(\omega)|$



$$\sin\left(\frac{\omega}{2}\right) = 0 \text{ when } \frac{\omega}{2} = \pi k \quad \omega = 2\pi k$$

$$\sin\left(\frac{\omega N}{2}\right) = 0 \text{ when } \frac{\omega N}{2} = \pi k \quad \omega = \frac{2\pi k}{N}$$

Generally, we'll plot & define $X(\omega)$ (DTFT) from $-\pi$ to π



periodic with period 2π

"Low" frequencies are centered on $2\pi k$, "high" on $\pi + 2\pi k$.